Cascaded and Hybrid Concepts for Voltage Conversion

Frederik Dostal  
Analog Devices, Inc.

There are different solutions for applications that require conversion from a high input voltage down to a very low output voltage. One interesting example is the conversion from 48 V down to 3.3 V. Such a specification is not only common in server applications for the information technology market, but in telecommunications as well.

![Figure 1](image1.png)  
**Figure 1.** Conversion of a voltage from 48 V down to 3.3 V in one single conversion step.

If a step-down converter (buck) is used for this single conversion step, as shown in Figure 1, the problem of small duty cycles emerges. The duty cycle is the relationship between the on-time (when the main switch is turned on) and the off-time (when the main switch is turned off). A buck converter has a duty cycle, which is defined by the following formula:

\[
\text{Duty Cycle} = \frac{\text{Output Voltage}}{\text{Input Voltage}}
\]

With an input voltage of 48 V and an output voltage of 3.3 V, the duty cycle is approximately 7%.

This means that at a switching frequency of 1 MHz (1000 ns per switching period), the Q1 switch is turned on for only 70 ns. Then, the Q1 switch is turned off for 930 ns and Q2 is turned on. For such a circuit, a switching regulator has to be chosen that allows for a minimum on-time of 70 ns or less. If such a component is selected, there is another challenge. Usually the very high power conversion efficiency of a buck regulator is reduced when operating at very short duty cycles. This is because there is only a very short time available to store energy in the inductor. The inductor needs to provide power for a long period during the off-time. This typically leads to very high peak currents in the circuit. To lower these currents, the inductance of L1 needs to be relatively large. This is due to the fact that during the on-time, a large voltage difference is applied across L1.

In the example, we see about 44.7 V across the inductor during the on-time, 48 V on the switch-node side, and 3.3 V on the output side. The inductor current is calculated by the following formula:

\[
i_L = \frac{1}{L} \int u_L dt
\]

If there is a high voltage across the inductor, the current rises during a fixed time period and at a fixed inductance. To reduce inductor peak currents, a higher inductance value needs to be selected. However, a higher value inductor adds to increased power losses. Under these voltage conditions, an efficient LTM8027 µModule regulator from Analog Devices achieves power efficiency of only 80% at 4 A output current.

Today, a very common and more efficient circuit solution to increase the power efficiency is the generation of an intermediate voltage. A cascaded setup with two highly efficient step-down (buck) regulators is shown in

![Figure 2](image2.png)  
**Figure 2.** Voltage conversion from 48 V down to 3.3 V in two steps, including a 12 V intermediate voltage.
For charge pumps, many developers assume a power output limitation of approximately 100 mW. The hybrid converter switch with the LTC7821 is designed for output currents of up to 25 A. For even higher performance, multiple LTC7821 controllers can be connected in a parallel multiphase configuration with synchronized frequency to share the overall load.

Figure 4. Typical conversion efficiency for converting 48 V to 5 V at a switching frequency of 500 kHz.

Figure 4 shows the typical conversion efficiency for a 48 V input voltage and a 5 V output voltage at different load currents. At approximately 6 A, a conversion efficiency exceeding 90% is reached. Between 13 A and 24 A, the efficiency is even higher than 94%.

A hybrid step-down controller supplies very high conversion efficiency in a compact form. It offers an interesting alternative to a discrete two-stage switching regulator design with intermediate bus voltage and to a single-stage converter that is forced to operate at a very low duty cycle. Some designers will prefer a cascaded architecture, others a hybrid architecture. With these two available options, every design should be successful.

About the Author
Frederik Dostal studied microelectronics at the University of Erlangen-Nuremberg, Germany. Starting work in the power management business in 2001, he has been active in various applications positions including four years in Phoenix, Arizona, working on switch mode power supplies. He joined Analog Devices in 2009 and works as a power management technical expert for Europe. He can be reached at frederik.dostal@analog.com.

Online Support Community
Engage with the Analog Devices technology experts in our online support community. Ask your tough design questions, browse FAQs, or join a conversation.

Visit ez.analog.com