

Rarely Asked Questions—Issue 122

The Truth About Voltage Feedback Resistors

By Tina Collins



Question:

The stability of my fully differential and voltage feedback amplifiers seems highly impacted by the value of my feedback resistors—the R_F/R_G ratio is always correct, so what's happening?

Answer:

When a signal needs gain, an amplifier is the component of choice. The ratio of the feedback and gain resistors, R_F/R_G , for a voltage feedback and a fully differential amplifier, determines the gain. Once the ratio is set, the next step is to select a value for either R_F or R_G . The choice of R_F can impact the stability of the amplifier.

An amplifier's internal input capacitance, found in the specification table of the data sheet, interacts with R_F to form a pole in the transfer function. If R_F is exceedingly large, this pole will affect stability. If the pole occurs at a frequency much larger than the crossover frequency, it will not affect stability. However, if the location of the pole as determined by $f = 1/(2\pi R_F C_{in,amp})$ occurs near the crossover frequency, the phase margin will be reduced leading to potential instability.

The example of Figure 1 shows the lab results of the small signal closed-loop gain vs. frequency response for the ADA4807-1 voltage feedback amplifier in a noninverting gain of 2 configuration with feedback resistors 499 Ω , 1 k Ω , and 10 k Ω . The data sheet recommended R_F value is 499 Ω .

The degree of peaking in the small signal frequency response indicates instability. Increasing R_F from 499 Ω to 1 k Ω marginally increases peaking. This would imply the amplifier has sufficient phase margin with an R_F of 1 k Ω and is stable. This is not the case for the R_F of 10 k Ω . The high level of peaking present implies instability (oscillation) and it is not recommended.

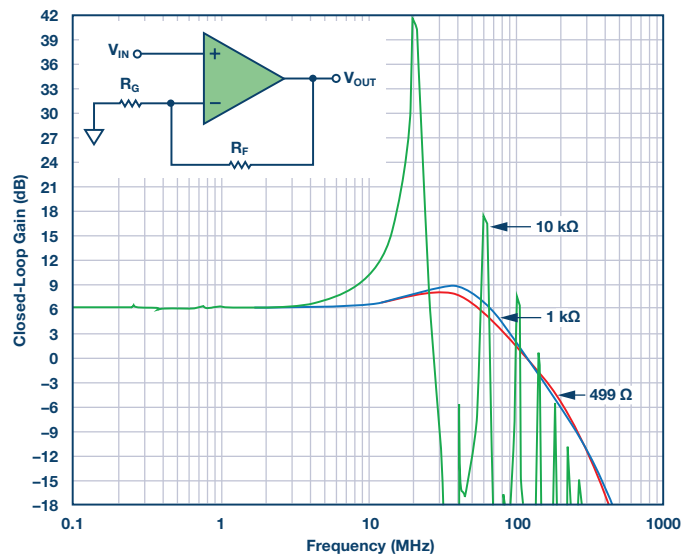


Figure 1. Lab results using different feedback resistors. $V_S = \pm 5 V$, $V_{OUT} = 40 mV p-p$, $R_{LOAD} = 1 k\Omega$ for R_F values of 499 Ω , 1 k Ω , and 10 k Ω .

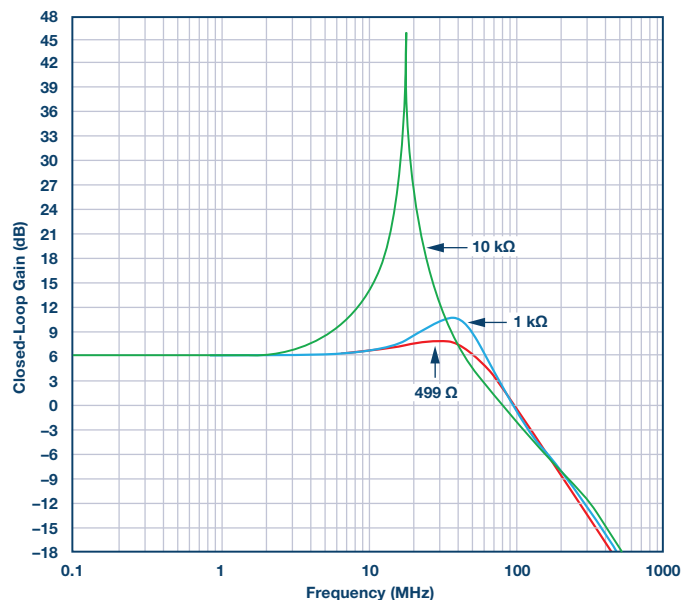


Figure 2. Simulation results using the ADA4807 SPICE model. $V_S = \pm 5 V$, $G = 2$ and $R_{LOAD} = 1 k\Omega$ for R_F values of 499 Ω , 1 k Ω , and 10 k Ω .

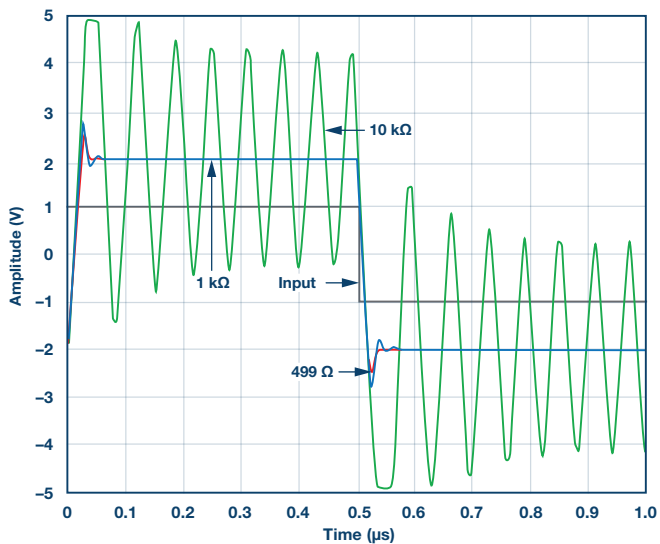


Figure 3. Pulse response simulation results using the ADA4807 SPICE model. $V_S = \pm 5\text{ V}$, $G = 2$ and $R_{LOAD} = 1\text{ k}\Omega$ for R_F values of $499\ \Omega$, $1\text{ k}\Omega$, and $10\text{ k}\Omega$.

Validating a circuit in the lab is not a mandatory step for verifying potential instabilities. Figure 3 shows the simulation results using the SPICE model with the same R_F values of $499\ \Omega$, $1\text{ k}\Omega$, and $10\text{ k}\Omega$. The results are consistent with Figure 1. Figure 3 shows the instability in the time domain. Adding a zero to the transfer function by placing a feedback capacitor across R_F will remove the instability as shown in Figure 4.

There are trade-offs in the selection of R_F , which are power dissipation, bandwidth, and stability. If power dissipation is critical, and the data sheet recommended feedback value cannot be used or a much higher R_F value is necessary, placing a feedback capacitor in parallel with R_F is an option. This choice results in lower bandwidth.

When selecting the R_F for a voltage feedback and a fully differential amplifier, consideration needs to be given to the system requirements. If speed is not critical, a feedback capacitor will help to stabilize a large R_F value. If speed is critical, the recommended data sheet R_F value is advised.

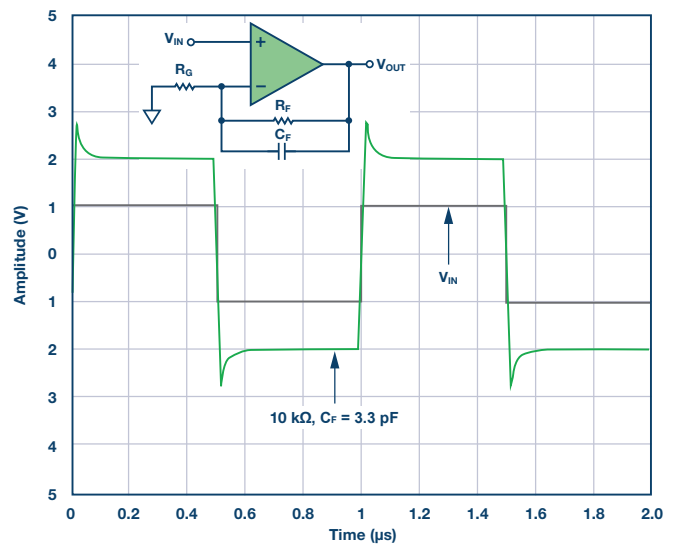


Figure 4. Pulse response simulation results with a 3.3 pF feedback capacitor, C_F . $V_S = \pm 5\text{ V}$, $G = 2$, $R_F = 10\text{ k}\Omega$ and $R_{LOAD} = 1\text{ k}\Omega$.

Ignoring the relationship of R_F with respect to stability, bandwidth, and power, can hinder a system or, worse yet, prevent the system from achieving its full performance.

References:

- Jung, Walter G. *Op Amp Applications Handbook*. Analog Devices, 2002.
- MT-033 Tutorial. [Voltage Feedback Op Amp Gain and Bandwidth](#).
- MT-044 Tutorial. [Op Amp Open-Loop Gain and Open-Loop Gain Nonlinearity](#).
- MT-059 Tutorial. [Compensating for the Effects of Input Capacitance on VFB and CFB Op Amps Used in Current-to-Voltage Converters](#).



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