

16-Bit, Isolated, Sigma-Delta ADC

FEATURES

- ▶ 5 MHz to 21 MHz main clock input frequency
- ▶ Offset drift vs. temperature: $\pm 0.6 \mu\text{V}/^\circ\text{C}$ maximum
- ▶ SNR: 86 dB typical
- ▶ 16 bits, no missing codes
- ▶ Full-scale analog input voltage range: $\pm 320 \text{ mV}$
- ▶ ENOB: 14 bits typical
- ▶ I_{DD1} : 10 mA maximum
- ▶ On-board digital isolator
- ▶ Operating temperature range
 - ▶ -40°C to $+125^\circ\text{C}$ (16-lead SOIC_W)
 - ▶ -40°C to $+105^\circ\text{C}$ (8-lead SOIC_IC)
- ▶ 20 V operation on V_{DD1}
- ▶ High CMTI: 150 kV/ μs minimum, $V_{\text{DD2}} = 3.3 \text{ V}$
- ▶ Wide-body SOICs
 - ▶ 16-lead SOIC_W
 - ▶ 8-lead SOIC_IC with increased creepage
- ▶ Safety and regulatory approvals
 - ▶ ADuM7703 wide-body, 16-lead SOIC_W, and ADuM7703 wide-body with increased creepage, 8-lead SOIC_IC
 - ▶ UL 1577
 - ▶ $V_{\text{ISO}} = 5700 \text{ V rms}$ for 1 minute
 - ▶ IEC/EN/CSA 62368-1
 - ▶ IEC/CSA 60601-1
 - ▶ IEC/CSA 61010-1
 - ▶ DIN EN IEC 60747-17 (VDE 0884-17)
 - ▶ $V_{\text{IORM}} = 1401 V_{\text{PEAK}}$

APPLICATIONS

- ▶ Shunt current monitoring
- ▶ AC motor controls
- ▶ Power and solar inverters
- ▶ Wind turbine inverters
- ▶ Analog-to-digital and optoisolator replacement

FUNCTIONAL BLOCK DIAGRAM

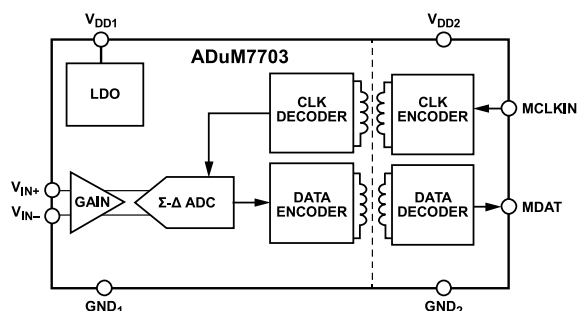


Figure 1.

GENERAL DESCRIPTION

The ADuM7703 is a high performance, second-order, Σ - Δ analog-to-digital converter (ADC) with an integrated low dropout (LDO) regulator that converts an analog input signal into a high speed, single-bit data stream with on-chip digital isolation based on Analog Devices, Inc., iCoupler® technology. The device operates from a 4.5 V to 20 V power supply range (V_{DD1}) and accepts a pseudo differential input signal of $\pm 250 \text{ mV}$ ($\pm 320 \text{ mV}$ full-scale). The pseudo differential input is ideally suited to shunt voltage monitoring in high voltage applications where galvanic isolation is required.

The analog input is continuously sampled by a high performance analog modulator and converted to a ones density digital output stream with a data rate of up to 21 MHz. The original information can be reconstructed with an appropriate sinc3 digital filter to achieve an 86 dB signal-to-noise ratio (SNR) at 78.1 kSPS with a 256 decimation rate and a 20 MHz main clock. The serial input and output operates from a 5 V or a 3.3 V supply (V_{DD2}).

The serial interface is digitally isolated. High speed complementary metal-oxide semiconductor (CMOS) technology, combined with monolithic transformer technology, results in the on-chip isolation providing outstanding performance characteristics, superior to alternatives such as optocoupler devices. The ADuM7703 device is available in a 16-lead, wide-body SOIC with an operating temperature range of -40°C to $+125^\circ\text{C}$ and an 8-lead, wide-body SOIC with an operating temperature range of -40°C to $+105^\circ\text{C}$.

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REVISION HISTORY**11/2024—Rev. A to Rev. B**

Changed Master to Main (Throughout).....	1
Changes to Features Section.....	1
Changes to Table 4.....	5
Changed Regulatory Information (Pending) Section to Regulatory Information Section.....	5
Changes to Regulatory Information Section.....	5
Changes to Table 5.....	5
Changed DIN V VDE V 0884-10 Insulation Characteristics Section to DIN EN IEC 60747-17 (VDE 0884-17) Insulation Characteristics Section	6
Changes to DIN EN IEC 60747-17 (VDE 0884-17) Insulation Characteristics Section, Table 6, and Figure 3 Caption.....	6
Deleted DIN V VDE V 0884-11 Insulation Characteristics (Pending) Section, Table 7, and Figure 4; Renumbered Sequentially.....	6
Change to Table 9.....	8

SPECIFICATIONS

V_{DD1} = 4.5 V to 20 V, V_{DD2} = 3 V or 5.5 V, V_{IN+} = -250 mV to +250 mV, V_{IN-} = 0 V, T_A = -40°C to +125°C (16-lead SOIC_W), T_A = -40°C to +105°C (8-lead SOIC_IC), MCLKIN frequency (f_{MCLKIN}) = 20 MHz, tested with a sinc3 filter, and a 256 decimation rate, unless otherwise noted.

Table 1.

Parameter	Min	Typ	Max	Unit	Test Conditions/Comments
STATIC PERFORMANCE					
Resolution	16			Bits	Filter output truncated to 16 bits
Integral Nonlinearity (INL) ¹		±2	±5	LSB	
Differential Nonlinearity (DNL) ¹			±0.99	LSB	Guaranteed no missed codes to 16 bits
Offset Error ¹		±0.05	±0.13	mV	Initial at T_A = 25°C
		±0.1	±0.18	mV	
Offset Drift vs. Temperature ¹		±0.1	±0.6	μV/°C	
Offset Drift vs. V_{DD1}		±0.1		μV/V	
Gain Error ¹			±0.2	% FSR	Initial at T_A = 25°C
Gain Error Drift vs. Temperature ¹		±12.5	±28	ppm/°C	
		±8	±18	μV/°C	
Gain Error Drift vs. V_{DD1}		±5		ppm/V	
ANALOG INPUT					
Input Voltage Range	-320		+320	mV	Full-scale range
	-250		+250	mV	For specified performance
Input Common-Mode Voltage Range		-0.2 to +0.8		V	
Dynamic Input Current		±1	±2	μA	V_{IN+} = ±250 mV, V_{IN-} = 0 V
		0.05		μA	V_{IN+} = 0 V, V_{IN-} = 0 V
DC Leakage Current		±0.01		μA	
Input Capacitance		25		pF	
DYNAMIC SPECIFICATIONS					
Signal-to-Noise-and-Distortion Ratio (SINAD) ¹	82	86		dB	V_{IN+} = 1 kHz
SNR ¹	83	86		dB	
Total Harmonic Distortion (THD) ¹	-84	-99		dB	
Peak Harmonic or Spurious-Free Dynamic Range Noise (SFDR) ¹		-97		dB	
Effective Number of Bits (ENOB) ¹	13.3	14		Bits	
ISOLATION COMMON-MODE TRANSIENT IMMUNITY (CMTI) ¹					
Static and Dynamic	75	150		kV/μs	Common-mode voltage ($ V_{CM} $) = 2 kV
	150			kV/μs	V_{DD2} = 5.5 V
					V_{DD2} = 3.3 V
LOGIC INPUTS					
Input High Voltage (V_{IH})	$0.7 \times V_{DD2}$			V	CMOS with Schmitt trigger
Input Low Voltage (V_{IL})			$0.3 \times V_{DD2}$	V	
Input Current (I_{IN})			±0.6	μA	
Input Capacitance (C_{IN})		10		pF	
LOGIC OUTPUTS					
Output High Voltage (V_{OH})	$V_{DD2} - 0.4$	$V_{DD2} - 0.2$		V	Output current (I_{OUT}) = -4 mA
Output Low Voltage (V_{OL})		0.2	0.4	V	I_{OUT} = 4 mA
POWER REQUIREMENTS					
V_{DD1}	4.5	15	20	V	V_{IN+} > 320 mV
V_{DD2}	3		5.5	V	
V_{DD1} Current (I_{DD1})		8.2	10	mA	
V_{DD2} Current (I_{DD2})		2	3	mA	
Power Dissipation		133	216.5	mW	V_{DD2} = 4.5 V to 5.5 V
		130	211	mW	V_{DD2} = 3 V to 3.6 V

SPECIFICATIONS

¹ See the [Terminology](#) section.

TIMING SPECIFICATIONS

V_{DD1} = 4.5 V to 20 V, V_{DD2} = 3 V or 5.5 V, T_A = -40°C to $+125^{\circ}\text{C}$ (16-lead SOIC_W), T_A = -40°C to $+105^{\circ}\text{C}$ (8-lead SOIC_IC), unless otherwise noted. Sample tested during initial release to ensure compliance. It is recommended to read MDAT on the MCLKIN rising edge.

Table 2.

Parameter	Limit at T_{MIN} , T_{MAX}			Unit	Description
	Min	Typ	Max		
f_{MCLKIN}	5	20	21	MHz	Main clock input frequency
t_{MCLKIN}	48	50	200	ns	Main clock input period
t_1 ¹			16	ns	Data access time after MCLKIN rising edge
t_2 ¹	5			ns	Data hold time after MCLKIN rising edge
t_3	$0.4 \times t_{MCLKIN}$			ns	Main clock low time
t_4	$0.4 \times t_{MCLKIN}$			ns	Main clock high time

¹ Defined as the time required from an 80% MCLKIN input level to when the output crosses $0.5 \times V_{DD2}$ as outlined in Figure 2. Measured with a $\pm 20 \mu\text{A}$ load and a 25 pF load capacitance.

Timing Diagram

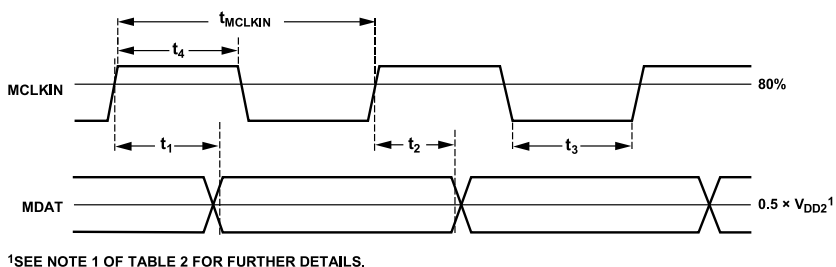


Figure 2. Data Timing Diagram

PACKAGE CHARACTERISTICS

Table 3.

Parameter ¹	Symbol	Min	Typ	Max	Unit	Test Conditions/Comments
Resistance (Input to Output)	R_{I-O}		10^{12}		Ω	
Capacitance (Input to Output)	C_{I-O}		1		pF	Frequency = 1 MHz

¹ The device is considered a 2-terminal device. For the 16-lead SOIC_W, Pin 1 to Pin 8 are shorted together and Pin 9 to Pin 16 are shorted together. For the 8-lead SOIC_IC, Pin 1 to Pin 4 are shorted together and Pin 5 to Pin 8 are shorted together.

SPECIFICATIONS

INSULATION AND SAFETY RELATED SPECIFICATIONS

Table 4.

Parameter	Symbol	Value	Unit	Test Conditions/Comments
Rated Dielectric Insulation Voltage	V _{ISO}	5700	V rms	1 minute duration
Minimum External Air Gap (Clearance) ^{1, 2}				
16-Lead SOIC_W	L(I01)	7.8	mm	Measured from input terminals to output terminals, shortest distance through air
8-Lead SOIC_IC	L(I01)	8.3	mm	Measured from input terminals to output terminals, shortest distance through air
Minimum External Tracking (Creepage) ¹				
16-Lead SOIC_W	L(I02)	7.8	mm	Measured from input terminals to output terminals, shortest distance path along body
8-Lead SOIC_IC	L(I02)	8.3	mm	Measured from input terminals to output terminals, shortest distance path along body
Minimum Internal Gap (Internal Clearance)		42	μm	Distance through insulation
Tracking Resistance (Comparative Tracking Index) ³	CTI	>600	V	DIN IEC 112/VDE 0303 Part 1
Material Group		I		Material Group (DIN VDE 0110, 1/89, Table I)

¹ In accordance with IEC 62368-1/IEC 60601-1 guidelines for the measurement of creepage and clearance distances for a pollution degree of 2 and altitudes ≤ 2000 m.

² Consideration must be given to pad layout to ensure the minimum required distance for clearance is maintained.

³ CTI rating for the ADuM7703 is >600 V and a Material Group I isolation group.

REGULATORY INFORMATION

The ADuM7703 certification approvals are listed in [Table 5](#).

Table 5. Safety Certifications

UL	CSA (Pending)	VDE
UL 1577 ¹ Single Protection, 5700 V rms	IEC/EN/CSA 62368-1 16-lead SOIC_W package Basic insulation, 780 V rms Reinforced insulation, 390 V rms 8-lead SOIC_IC package Basic insulation, 830 V rms Reinforced insulation, 415 V rms IEC/CSA 61010-1 Basic insulation, 600 V rms Reinforced insulation, 300 V rms IEC/CSA 60601-1 16-lead SOIC_W package Reinforced insulation (2 MOPP), 237.5 V rms 8-lead SOIC_IC package Reinforced insulation (2 MOPP), 261 V rms	DIN EN IEC 60747-17 (VDE 0884-17) ² Reinforced insulation, 1401 V _{PEAK}
File E214100	File No: 205078	Certificate No: 40011599

¹ In accordance with UL 1577, each ADuM7703 is proof tested by applying an insulation test voltage ≥ 6840 V rms for 1 sec (current leakage detection limit = 15 μA).

² In accordance with DIN EN IEC 60747-17 (VDE 0884-17), each ADuM7703 is proof tested by applying an insulation test voltage ≥ 2627 V_{PEAK} for 1 sec (partial discharge detection limit = 5 pC).

SPECIFICATIONS

DIN EN IEC 60747-17 (VDE 0884-17) INSULATION CHARACTERISTICS

This isolator is suitable for reinforced electrical isolation only within the safety limit data. Maintenance of the safety data is ensured by means of protective circuits. The asterisk (*) marking on packages denotes DIN EN IEC 60747-17 (VDE 0884-17) approval.

Table 6.

Description	Symbol	Characteristic	Unit
INSTALLATION CLASSIFICATION PER DIN VDE 0110			
For Rated Mains Voltage ≤ 300 V rms		I to IV	
For Rated Mains Voltage ≤ 450 V rms		I to IV	
For Rated Mains Voltage ≤ 600 V rms		I to IV	
CLIMATIC CLASSIFICATION		40/125/21	
POLLUTION DEGREE (DIN VDE 0110, TABLE 1)		2	
MAXIMUM REPETITIVE ISOLATION VOLTAGE	V_{IORM}	1401	V_{PEAK}
MAXIMUM WORKING ISOLATION VOLTAGE	V_{IOWM}	991	V rms
INPUT TO OUTPUT TEST VOLTAGE, METHOD B1 $V_{IORM} \times 1.875 = V_{PR}$, 100% Production Test, $t_m = 1$ Second, Partial Discharge < 5 pC	$V_{PD(M)}$	2627	V_{PEAK}
INPUT TO OUTPUT TEST VOLTAGE, METHOD A	$V_{PR(M)}$		
After Environmental Test Subgroup 1 $V_{IORM} \times 1.6 = V_{PR}$, $t_m = 60$ sec, Partial Discharge < 5 pC		2242	V_{PEAK}
After Input and/or Safety Test Subgroup 2/Safety Test Subgroup 3 $V_{IORM} \times 1.2 = V_{PR}$, $t_m = 60$ sec, Partial Discharge < 5 pC		1681	V_{PEAK}
MAXIMUM TRANSIENT ISOLATION VOLTAGE $V_{TEST} = 1.2 \times V_{IOTM}$, $t = 1$ s (100% Production)	V_{IOTM}	8000	V_{PEAK}
MAXIMUM IMPULSE VOLTAGE Surge Voltage in Air, Waveform per IEC 61000-4-5	V_{IMP}	8000	V_{PEAK}
MAXIMUM SURGE ISOLATION VOLTAGE $V_{TEST} \geq 1.3 \times V_{IMP}$ (Sample Test), Tested in Oil, Waveform per IEC 61000-4-5	V_{IOSM}	12800	V_{PEAK}
SAFETY LIMITING VALUES (MAXIMUM VALUE ALLOWED IN THE EVENT OF A FAILURE) ¹			
Case Temperature	T_S	150	°C
Side 1 (P_{VDD1}) and Side 2 (P_{VDD2}) Power Dissipation	P_{SO}		
16-Lead SOIC_W		1.43	W
8-Lead SOIC_IC		1.19	W
INSULATION RESISTANCE AT T_S , VOLTAGE INPUT TO OUTPUT (V_{IO}) = 500 V	R_{IO}	$> 10^9$	Ω

¹ See Figure 3.

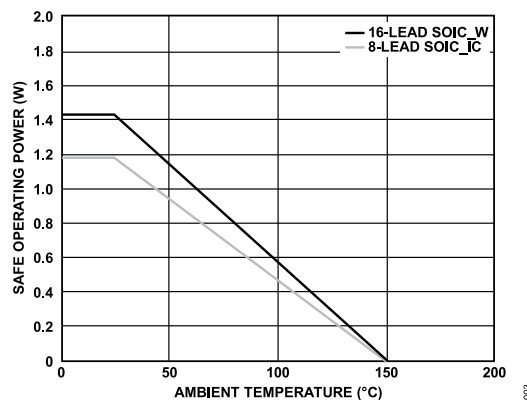


Figure 3. Thermal Derating Curve, Dependence of Safety Limiting Values with Case Temperature per DIN EN IEC 60747-17 (VDE 0884-17)

ABSOLUTE MAXIMUM RATINGS

$T_A = 25^\circ\text{C}$, unless otherwise noted. All voltages are relative to the respective GND_x .

Table 7.

Parameter	Rating
V_{DD1} to GND_1	$-0.3\text{ V to }+23\text{ V}$
V_{DD2} to GND_2	$-0.3\text{ V to }+6\text{ V}$
Analog Input Voltage to GND_1	$-1\text{ V to }+4.3\text{ V}$
Digital Input Voltage to GND_2	$-0.5\text{ V to }V_{DD2} + 0.5\text{ V}$
Digital Output Voltage to GND_2	$-0.5\text{ V to }V_{DD2} + 0.5\text{ V}$
Input Current to Any Pin Except Supplies ¹	$\pm 10\text{ mA}$
Output Current from Any Pin Except Supplies	$\pm 10\text{ mA}$
Operating Temperature Range	
16-lead SOIC_W	$-40^\circ\text{C to }+125^\circ\text{C}$
8-lead SOIC_IC	$-40^\circ\text{C to }+105^\circ\text{C}$
Storage Temperature Range	$-65^\circ\text{C to }+150^\circ\text{C}$
Junction Temperature	150°C
Pb-Free Temperature, Soldering	
Reflow	260°C
Electrostatic Discharge (ESD)	
Field Induced Charged Device Model (FICDM) ²	$\pm 1250\text{ V}$
Human Body Model (HBM) ³	$\pm 4000\text{ V}$

¹ Transient currents of up to 100 mA do not cause silicon controlled rectifier (SCR) to latch up.

² JESD22-C101, resistor, capacitor (RC) network, 1 Ω , package capacitance, and Class IV.

³ ESDA/JEDEC JS-001-2011, RC network: 1.5 k Ω , 100 pF, and Class 3A.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Close attention to PCB thermal design is required.

Table 8. Thermal Resistance

Package Type ¹	θ_{JA} ²	Ψ_{JC} ³	Unit
RI-8-1	105	9.25	$^\circ\text{C/W}$
RW-16	87.25	10.4	$^\circ\text{C/W}$

¹ Thermal impedance simulated values are based on a JEDEC 2S2P thermal test board. See JEDEC JESD-51.

² θ_{JA} was calculated using the total power and maximum junction temperature.

³ Ψ_{JC} was calculated using the package center case temperature.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

ABSOLUTE MAXIMUM RATINGS

INSULATION RATINGS

The maximum continuous working voltage refers to the continuous voltage magnitude imposed across the isolation barrier. See the [Insulation Lifetime](#) section for more details.

Table 9. Maximum Continuous Working Voltage

Parameter	Insulation Rating ¹	Lifetime Conditions
Basic Insulation		
AC Voltage		
Bipolar Waveform	1129 V _{PEAK}	20 years to 1000 ppm failure at 1129 V _{PEAK} (798 V rms, 50 Hz or 60 Hz sine wave)
Reinforced Insulation		
AC Voltage		
Bipolar Waveform	1060 V _{PEAK}	20 years to 1 ppm failure at 1060 V _{PEAK} (750 V rms, 50 Hz or 60 Hz sine wave)

¹ Insulation capability without regard to creepage limitations. Working voltage may be limited by the PCB creepage when considering rms voltages for components soldered to a PCB (assumes Material Group I up to 1270 V rms), or package: RI-8-1 package creepage of 8.3 mm, and RW-16 package creepage of 7.8 mm, when considering rms voltages for Material Group I.

PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

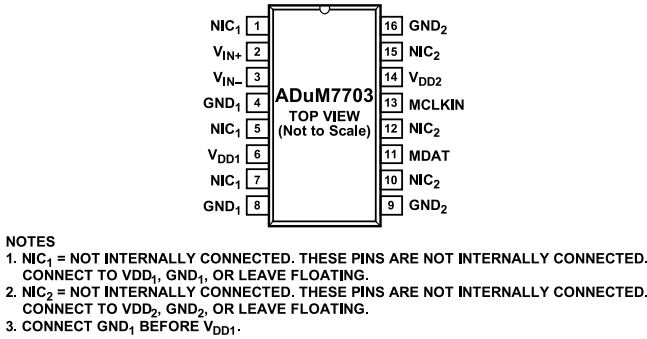
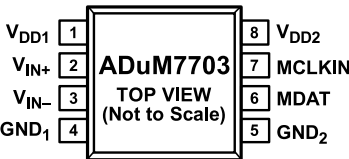


Figure 4. 16-Lead SOIC_W Pin Configuration

Table 10. 16-Lead SOIC_W Pin Function Descriptions

Pin No.	Mnemonic	Description
1, 5, 7	NIC_1	Not Internally Connected. These pins are not internally connected. Connect these pins to V_{DD1} , GND_1 , or leave floating.
2	V_{IN+}	Positive Analog Input.
3	V_{IN-}	Negative Analog Input.
4, 8	GND_1	Ground 1. These pins are the ground reference point for all circuitry on the isolated side.
6	V_{DD1}	Supply Voltage, 4.5 V to 20 V. This pin is the supply voltage for the isolated side of the ADuM7703 and is relative to GND_1 . For device operation, connect the supply voltage to Pin 7. Decouple the supply pin to GND_1 with a 10 μ F capacitor in parallel with a 100 nF capacitor as close to the pin as possible.
9, 16	GND_2	Ground 2. These pins are the ground reference point for all circuitry on the nonisolated side.
10, 12, 15	NIC_2	Not Internally Connected. These pins are not internally connected. Connect these pins to V_{DD2} , GND_2 , or leave floating.
11	MDAT	Serial Data Output. The single-bit modulator output is supplied to this pin as a serial data stream. The bits are clocked out on the rising edge of the MCLKIN input and are valid on the following MCLKIN rising edge.
13	MCLKIN	Main Clock Logic Input, 5 MHz to 21 MHz frequency range. The bit stream from the modulator is propagated on the rising edge of the MCLKIN.
14	V_{DD2}	Supply Voltage, 3 V or 5.5 V. This pin is the supply voltage for the nonisolated side and is relative to GND_2 . Decouple this supply to GND_2 with a 10 μ F capacitor in parallel with a 100 nF capacitor as close to the pin as possible.

PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS



NOTES
1. CONNECT GND₁ BEFORE V_{DD1}.

Figure 5. 8-Lead SOIC_IC Pin Configuration

Table 11. 8-Lead SOIC_IC Pin Function Descriptions

Pin No.	Mnemonic	Description
1	V _{DD1}	Supply Voltage, 4.5 V to 20 V. This pin is the supply voltage for the isolated side of the ADuM7703 and is relative to GND ₁ . For device operation, connect the V _{DD1} supply voltage to the V _{DD1} pin. Decouple the supply pin to GND ₁ with a 10 μ F capacitor in parallel with a 100 nF capacitor as close to the pin as possible.
2	V _{IN+}	Positive Analog Input.
3	V _{IN-}	Negative Analog Input.
4	GND ₁	Ground 1. This pin is the ground reference point for all circuitry on the isolated side.
5	GND ₂	Ground 2. This pin is the ground reference point for all circuitry on the nonisolated side.
6	MDAT	Serial Data Output. The single-bit modulator output is supplied to this pin as a serial data stream. The bits are clocked out on the rising edge of the MCLKIN input and are valid on the following MCLKIN rising edge.
7	MCLKIN	Main Clock Logic Input, 5 MHz to 21 MHz frequency range. The bit stream from the modulator is propagated on the rising edge of the MCLKIN.
8	V _{DD2}	Supply Voltage, 3 V or 5.5 V. This pin is the supply voltage for the nonisolated side and is relative to GND ₂ . Decouple this supply to GND ₂ with a 10 μ F capacitor in parallel with a 100 nF capacitor as close to the pin as possible.

TYPICAL PERFORMANCE CHARACTERISTICS

$T_A = 25^\circ\text{C}$, $V_{DD1} = 5\text{ V}$, $V_{DD2} = 5\text{ V}$, $V_{IN+} = -250\text{ mV}$ to $+250\text{ mV}$, $V_{IN-} = 0\text{ V}$, and $f_{MCLKIN} = 20\text{ MHz}$, using a sinc3 filter with a 256 oversampling ratio (OSR), unless otherwise noted.

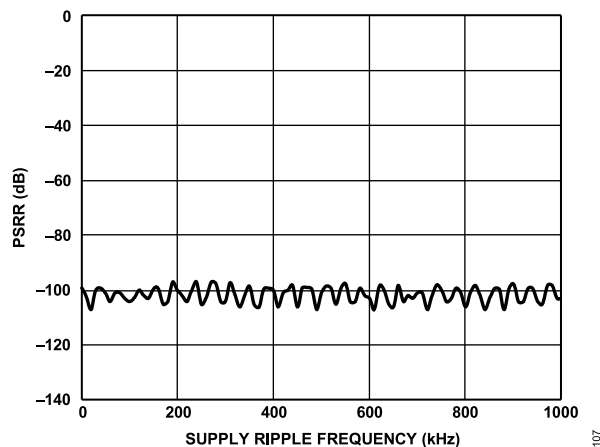


Figure 6. Power Supply Rejection Ratio (PSRR) vs. Supply Ripple Frequency

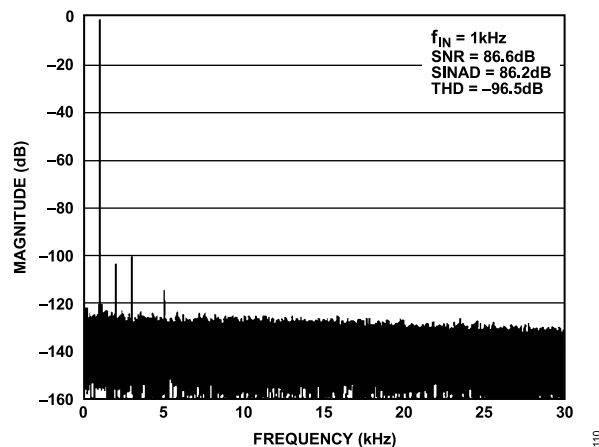


Figure 9. Typical Fast Fourier Transform (FFT) (f_{IN} is the Input Frequency)

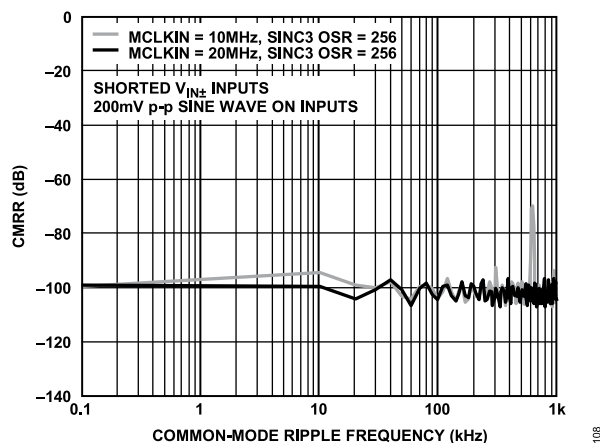


Figure 7. Common-Mode Rejection Ratio (CMRR) vs. Common-Mode Ripple Frequency

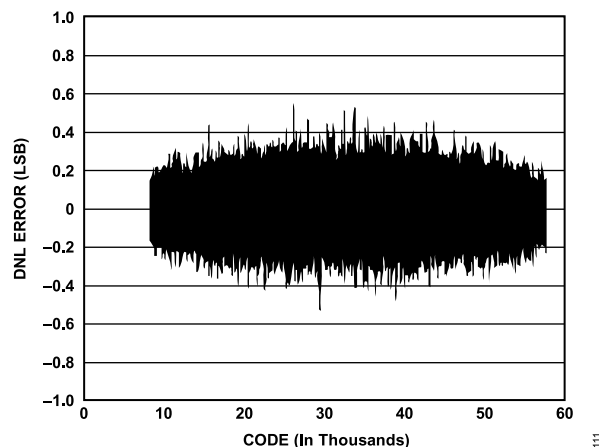


Figure 10. Typical DNL Error

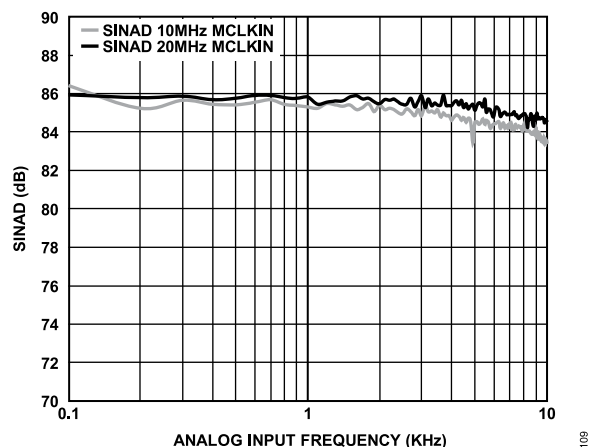


Figure 8. SINAD vs. Analog Input Frequency

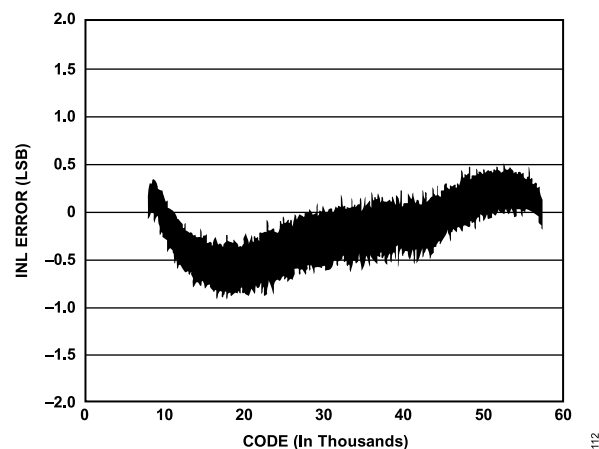


Figure 11. Typical INL Error

TYPICAL PERFORMANCE CHARACTERISTICS

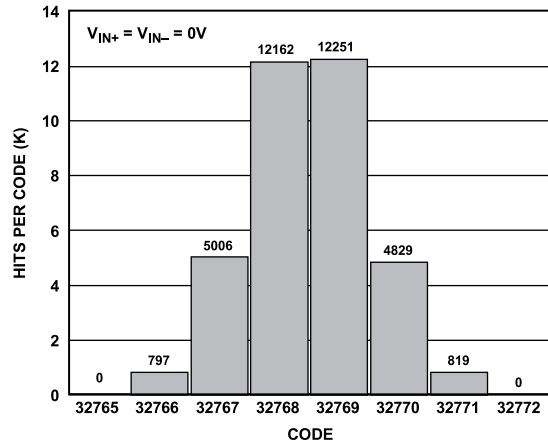


Figure 12. Histogram of Codes at the Code Center

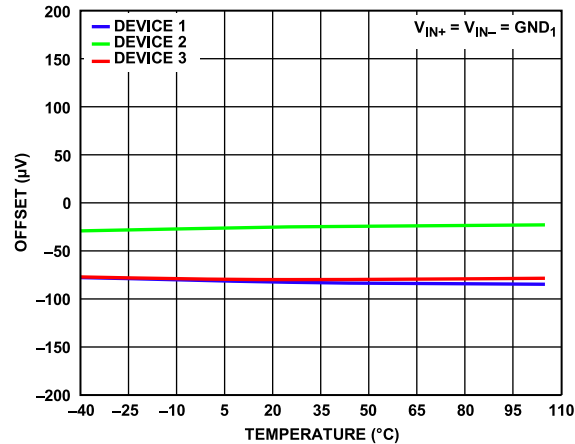


Figure 15. Offset vs. Temperature

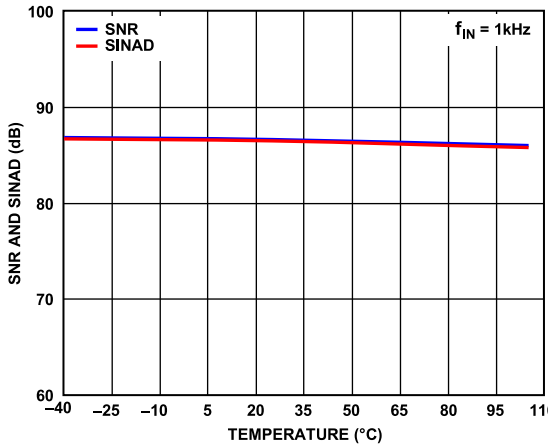


Figure 13. SNR and SINAD vs. Temperature

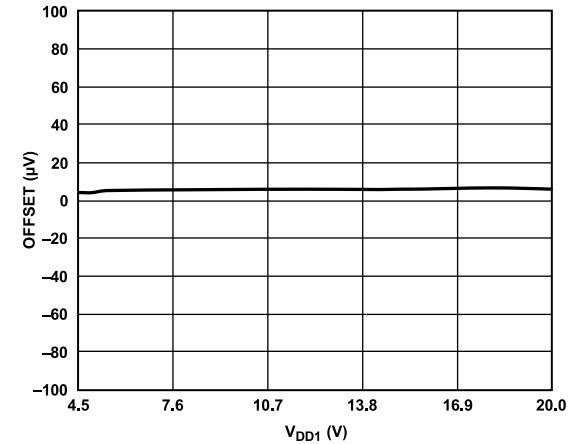


Figure 16. Offset vs. VDD1

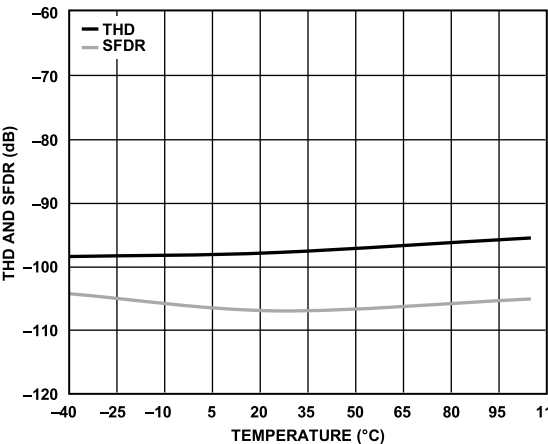


Figure 14. THD and SFDR vs. Temperature

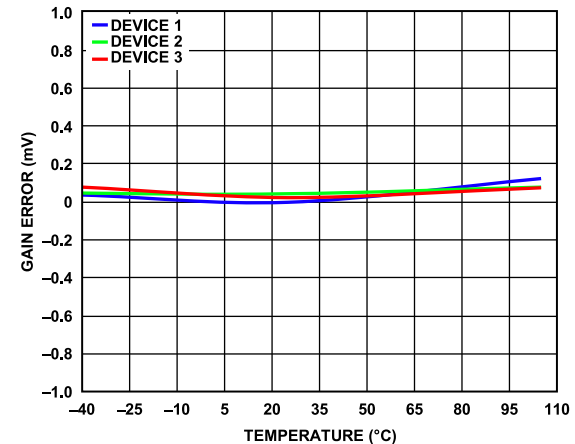
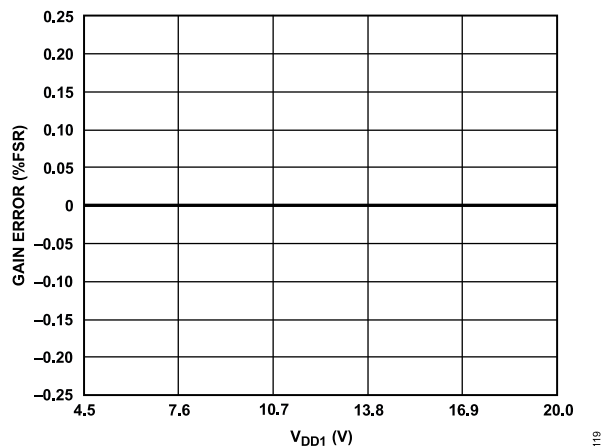
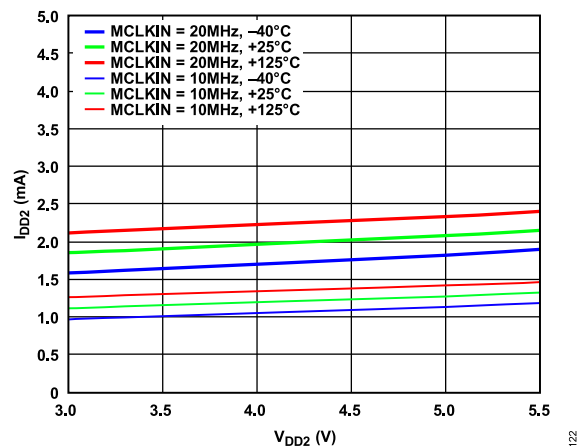
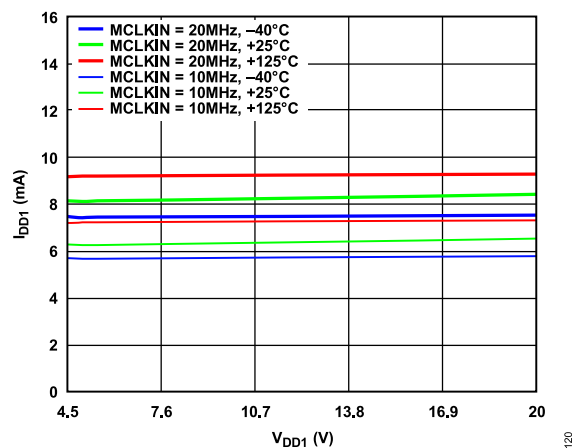
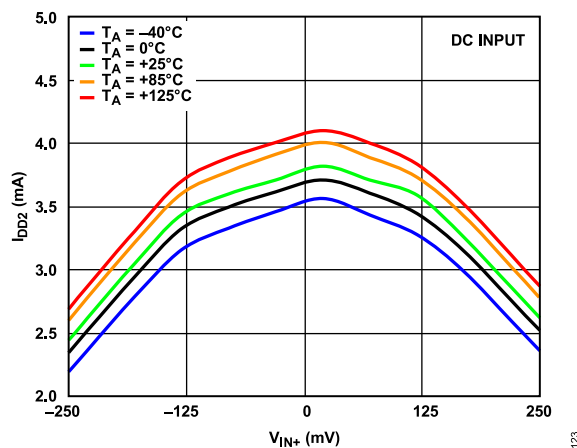
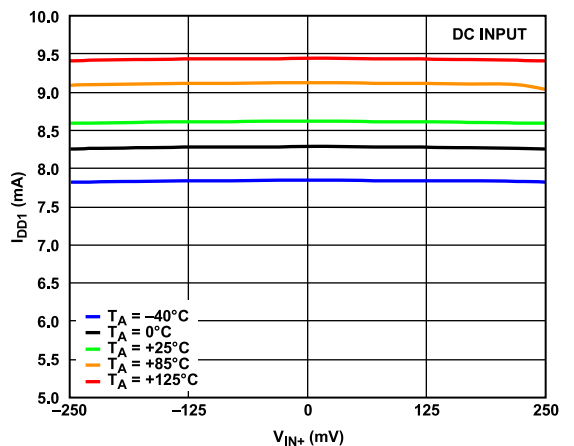
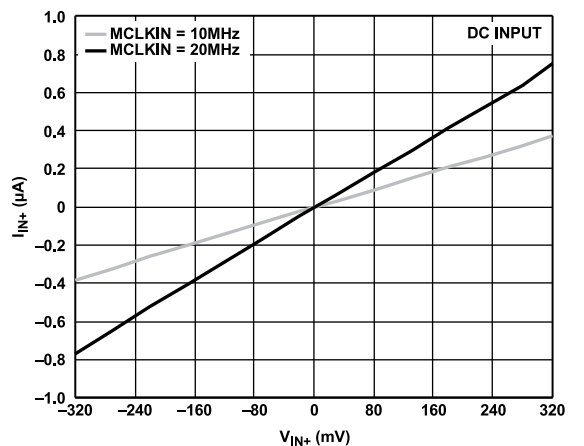


Figure 17. Gain Error vs. Temperature

TYPICAL PERFORMANCE CHARACTERISTICS

Figure 18. Gain Error vs. V_{DD1} Figure 21. I_{DD2} vs. V_{DD2} at Various Clock Rates and TemperaturesFigure 19. I_{DD1} vs. V_{DD1} at Various Clock Rates and TemperaturesFigure 22. I_{DD2} vs. V_{IN+} DC Input at Various TemperaturesFigure 20. I_{DD1} vs. V_{IN+} DC Input at Various TemperaturesFigure 23. V_{IN+} Current (I_{IN+}) vs. V_{IN+} DC Input

TERMINOLOGY

Differential Nonlinearity (DNL)

DNL is the difference between the measured and the ideal 1 LSB change between any two adjacent codes in the analog-to-digital converter (ADC).

Integral Nonlinearity (INL)

INL is the maximum deviation from a straight line passing through the endpoints of the ADC transfer function. The endpoints of the transfer function are specified negative full-scale, -250 mV ($V_{IN+} - V_{IN-}$), Code 7168 for the 16-bit level, and specified positive full scale, +250 mV ($V_{IN+} - V_{IN-}$), Code 58,368 for the 16-bit level.

Offset Error

Offset error is the deviation of the midscale code (32,768 for the 16-bit level) from the ideal $V_{IN+} - V_{IN-}$ (that is, 0 V).

Offset Drift vs. Temperature

The offset drift is calculated using the box method, as shown by the following equation:

$$\text{Offset Drift} = ((\text{Voltage}_{MAX} - \text{Voltage}_{MIN})/T_{\Delta}) \quad (1)$$

where:

Voltage_{MAX} is the maximum offset error point recorded.

Voltage_{MIN} is the minimum offset error point recorded.

T_{Δ} is the difference in temperature between the maximum and minimum operating range.

Gain Error

The gain error includes both the positive full-scale gain error and the negative full-scale gain error. The positive full-scale gain error is the deviation of the specified positive full-scale code (58,368 for the 16-bit level) from the ideal $V_{IN+} - V_{IN-}$ (250 mV) after the offset error is adjusted out. The negative full-scale gain error is the deviation of the specified negative full-scale code (7168 for the 16-bit level) from the ideal $V_{IN+} - V_{IN-}$ (-250 mV) after the offset error is adjusted out.

Gain Error Drift vs. Temperature

The gain error drift (GED) is calculated using the box method, as shown by the following equation:

$$\text{GED (ppm)} = ((\text{Voltage}_{MAX} - \text{Voltage}_{MIN})/(\text{Voltage}_{FS} \times T_{\Delta})) \times 10^6 \quad (2)$$

where:

Voltage_{MAX} is the maximum gain error point recorded.

Voltage_{MIN} is the minimum gain error point recorded.

Voltage_{FS} is the analog input range full scale.

T_{Δ} is the difference in temperature between the maximum and minimum operating range.

Signal-to-Noise-and-Distortion Ratio (SINAD)

SINAD is the measured ratio of signal to noise and distortion at the output of the ADC. The signal is the rms value of the sine wave, and noise is the rms sum of all nonfundamental signals up to half the sampling frequency ($f_S/2$), including harmonics, but excluding dc.

Signal-to-Noise Ratio (SNR)

SNR is the measured ratio of signal to noise at the output of the ADC. The signal is the rms amplitude of the fundamental. Noise is the sum of all nonfundamental signals up to half the sampling frequency ($f_S/2$), excluding dc.

The ratio is dependent on the number of quantization levels in the digitization process: the greater the number of levels, the smaller the quantization noise. The theoretical SNR for an ideal N-bit converter with a sine wave input is given by

$$\text{SNR} = (6.02N + 1.76) \text{ dB} \quad (3)$$

Therefore, for a 12-bit converter, the SNR is 74 dB.

Isolation Common-Mode Transient Immunity (CMTI)

The isolation CMTI specifies the rate of the rise and fall of a transient pulse applied across the isolation boundary, beyond which clock or data is corrupted. Both the rate of change and the absolute common-mode voltage of the pulse are recorded. The ADuM7703 is tested under both static and dynamic CMTI conditions. Static testing detects single-bit errors from the device. Dynamic testing monitors the filtered data output for variations in noise performance to a randomized application of the CMTI pulse.

Total Harmonic Distortion (THD)

THD is the ratio of the rms sum of the harmonics to the fundamental. It is defined as

$$\text{THD (dB)} = 20 \log \frac{\sqrt{V_2^2 + V_3^2 + V_4^2 + V_5^2 + V_6^2}}{V_1} \quad (4)$$

where:

V_1 is the rms amplitude of the fundamental.

V_2 , V_3 , V_4 , V_5 , and V_6 are the rms amplitudes of the second through the sixth harmonics.

Peak Harmonic or Spurious-Free Dynamic Range (SFDR) Noise

Peak harmonic or SFDR noise is defined as the ratio of the rms value of the next largest component in the ADC output spectrum (up to $f_S/2$, excluding dc) to the rms value of the fundamental. Normally, the value of this specification is determined by the largest harmonic in the spectrum, but for ADCs where the harmonics are buried in the noise floor, it is a noise peak.

TERMINOLOGY

Effective Number of Bits (ENOB)

ENOB is defined by

$$ENOB = (SINAD - 1.76)/6.02 \text{ bits} \quad (5)$$

Noise Free Code Resolution

Noise free code resolution represents the resolution in bits for which there is no code flicker. The noise free code resolution for an N-bit converter is defined as

$$\text{Noise Free Code Resolution (Bits)} = \log_2(2^N / \text{Peak-to-Peak Noise}) \quad (6)$$

The peak-to-peak noise in LSBs is measured with $V_{IN+} = V_{IN-} = 0 \text{ V}$.

Common-Mode Rejection Ratio (CMRR)

CMRR is the ratio of the power in the ADC output at $\pm 250 \text{ mV}$ frequency, f , to the power of a $\pm 250 \text{ mV}$ p-p sine wave applied to the common-mode voltage of V_{IN+} and V_{IN-} of frequency, f_S , as

$$CMRR \text{ (dB)} = 10 \log(P_f/P_{f_S}) \quad (7)$$

where:

P_f is the power at frequency, f , in the ADC output.

P_{f_S} is the power at frequency, f_S , in the ADC output.

Power Supply Rejection Ratio (PSRR)

Variations in power supply affect the full-scale transition but not the linearity of the converter. PSRR is the maximum change in the specified full-scale ($\pm 250 \text{ mV}$) transition point due to a change in power supply voltage from the nominal value.

THEORY OF OPERATION

CIRCUIT INFORMATION

The ADuM7703 isolated $\Sigma\Delta$ modulator converts an analog input signal to a high speed (21 MHz maximum), single-bit data stream. The time average single-bit data from the modulator is directly proportional to the input signal. [Figure 24](#) shows a typical application circuit where the ADuM7703 provides isolation between the analog input, a current sensing resistor or shunt, and the digital output, which is then processed by a digital filter to provide an N-bit word.

ANALOG INPUT

The pseudo differential analog input of the ADuM7703 is implemented with a switched capacitor circuit. This circuit implements a second-order modulator stage that digitizes the input signal to a single-bit output stream. The sample clock (MCLKIN) provides the clock signal for the conversion process as well as the output data framing clock. This clock source is externally supplied to the ADuM7703. The analog input signal is continuously sampled by the modulator and compared to an internal voltage reference. A digital stream that accurately represents the analog input over time appears at the output of the converter (see [Figure 25](#)).

A differential signal of 0 mV ideally results in a stream of alternating 1s and 0s at the MDAT output pin. This output is high 50% of the time and low 50% of the time. A differential input of 250 mV produces a stream of 1s and 0s that are high 89.06% of the time. A differential input of -250 mV produces a stream of 1s and 0s that are high 10.94% of the time.

A differential input of 320 mV ideally results in a stream of all 1s. A differential input of -320 mV ideally results in a stream of all 0s. The ADuM7703 absolute full-scale range is ± 320 mV, and the specified full-scale performance range is ± 250 mV, as shown in [Table 12](#).

Table 12. Analog Input Range

Analog Input	Voltage Input (mV)
Positive Full-Scale (+FS) Value	+320
Positive Specified Performance	+250
Zero	0
Negative Specified Performance	-250
Negative Full-Scale (-FS) Value	-320

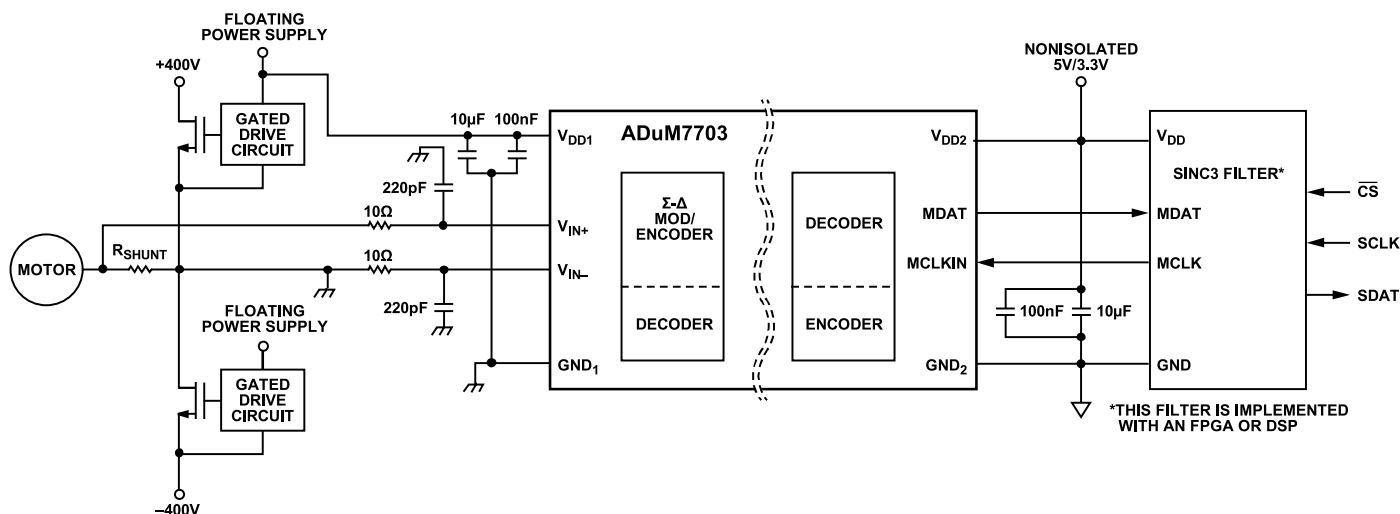


Figure 24. Typical Application Circuit

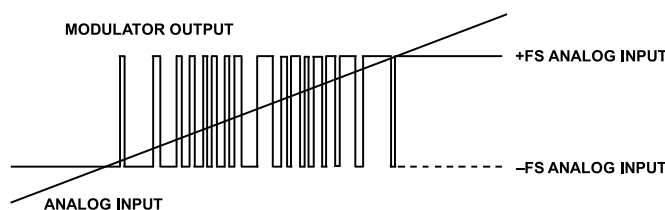


Figure 25. Analog Input vs. Modulator Output

THEORY OF OPERATION

To reconstruct the original information, this output must be digitally filtered and decimated. A sinc3 filter is recommended because this filter is one order higher than that of the ADuM7703 modulator, which is a second-order modulator. When a 256 decimation rate is used, the resulting 16-bit word rate is 78.1 kSPS, assuming a

20 MHz external clock frequency. See the [Digital Filter](#) section for more detailed information on the sinc filter implementation. [Figure 26](#) shows the transfer function of the ADuM7703 relative to the 16-bit output.

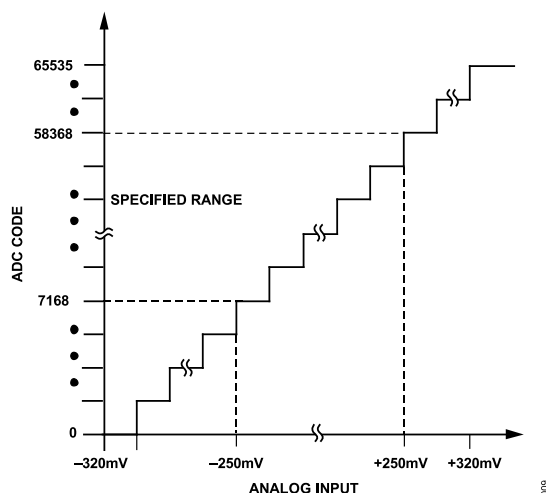


Figure 26. Filtered and Decimated 16-Bit Transfer Function

APPLICATIONS INFORMATION

CURRENT SENSING APPLICATIONS

The ADuM7703 is ideally suited for current sensing applications where the voltage across a shunt resistor (R_{SHUNT}) is monitored. The load current flowing through an external shunt resistor produces a voltage at the input terminals of the ADuM7703. The ADuM7703 provides isolation between the analog input from the current sensing resistor and the digital outputs. By selecting the appropriate shunt resistor value, a variety of current ranges can be monitored.

Choosing R_{SHUNT}

The shunt resistor (R_{SHUNT}) values used in conjunction with the ADuM7703 are determined by the specific application requirements in terms of voltage, current, and power. Small resistors minimize power dissipation, whereas low inductance resistors prevent any induced voltage spikes, and high tolerance devices reduce current variations. The final values chosen are a compromise between low power dissipation and accuracy. Higher value resistors use the full performance input range of the ADC, thus achieving maximum SNR performance. Low value resistors dissipate less power but do not use the full performance input range. The ADuM7703, however, delivers excellent performance, even with lower input signal levels, allowing low value shunt resistors to be used while maintaining system performance.

To choose a suitable shunt resistor, first determine the current through the shunt. Calculated the shunt current for a 3-phase induction motor as

$$I_{RMS} = PW / (1.73 \times V \times EF \times PF) \quad (8)$$

where:

I_{RMS} is the motor phase current (A rms).

PW is the motor power (W).

V is the motor supply voltage (V ac).

EF is the motor efficiency (%).

PF is the power efficiency (%).

To determine the shunt peak sense current (I_{SENSE}), consider the motor phase current and any overload that may be possible in the system. When the peak sense current is known, divide the voltage range of the ADuM7703 (± 250 mV) by the peak sense current to yield a maximum shunt value.

If the power dissipation in the shunt resistor is too large, the shunt resistor can be reduced and less of the ADC input range can be used. Figure 27 shows the SINAD performance characteristics and the ENOB of resolution for the ADuM7703 for different input signal amplitudes. The performance of the ADuM7703 at lower input signal ranges allows smaller shunt values to be used while still maintaining a high level of performance and overall system efficiency.

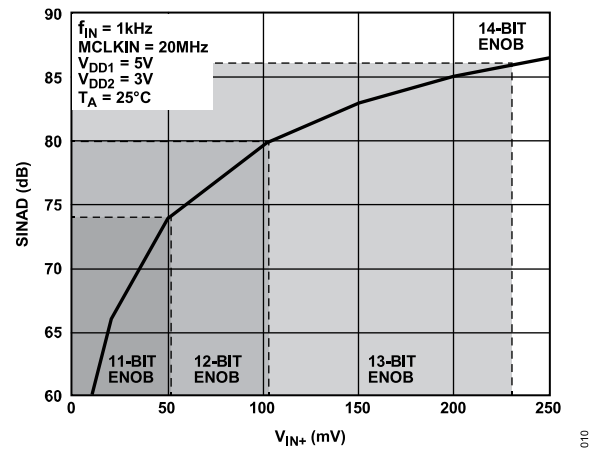


Figure 27. SINAD vs. V_{IN+} AC Input Signal Amplitude

R_{SHUNT} must dissipate the current² × resistance (I^2R) power losses. If the power dissipation rating of the resistor is exceeded, the value may drift, or the resistor may be damaged, resulting in an open circuit. This open circuit can result in a differential voltage across the terminals of the ADuM7703, in excess of the absolute maximum ratings. If I_{SENSE} has a large high frequency component, choose a resistor with low inductance.

VOLTAGE SENSING APPLICATIONS

The ADuM7703 can also be used for isolated voltage monitoring. For example, in motor control applications, the device can be used to sense the bus voltage. In applications where the voltage being monitored exceeds the specified analog input range of the ADuM7703, a voltage divider network can be used to reduce the voltage being monitored to the required range.

INPUT FILTER

In a typical use case for directly measuring the voltage across a shunt resistor, the ADuM7703 can be connected directly across the shunt resistor with a simple RC low-pass filter on each input.

The recommended circuit configuration for driving the differential inputs to achieve best performance is shown in Figure 28. An RC low-pass filter is placed on both the analog input pins. Recommended values for the resistors and capacitors are 10 Ω and 220 pF, respectively. If possible, equalize the source impedance on each analog input to minimize offset.

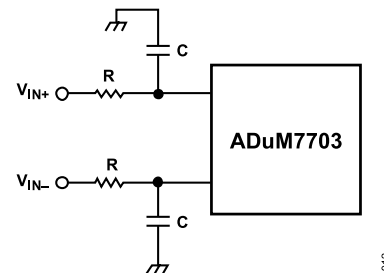


Figure 28. RC Low-Pass Filter Input Network

APPLICATIONS INFORMATION

The input filter configuration for the ADuM7703 is not limited to the low-pass structure shown in Figure 28. The differential RC filter configuration shown in Figure 29 also achieves excellent performance. Recommended values for the resistors and capacitor are 22 Ω and 47 pF, respectively.

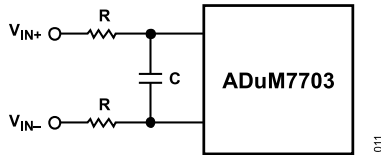


Figure 29. Differential RC Filter Network

DIGITAL FILTER

The output of the ADuM7703 is a continuous digital bit stream. To reconstruct the original input signal information, this output bit stream must be digitally filtered and decimated. A sinc filter is recommended due to simplicity of the filter. A sinc3 filter is recommended because the filter is one order higher than that of the ADuM7703 modulator, which is a second-order modulator. The type of filter selected, the decimation rate, and the modulator clock used determines the overall system resolution and throughput rate. The higher the decimation rate, the greater the system accuracy, as shown in Figure 30. However, there is a trade-off between accuracy and throughput rate and, therefore, higher decimation rates result in lower throughput solutions. Note that for a given bandwidth requirement, a higher MCLKIN frequency can allow higher decimation rates to be used, resulting in higher SNR performance.

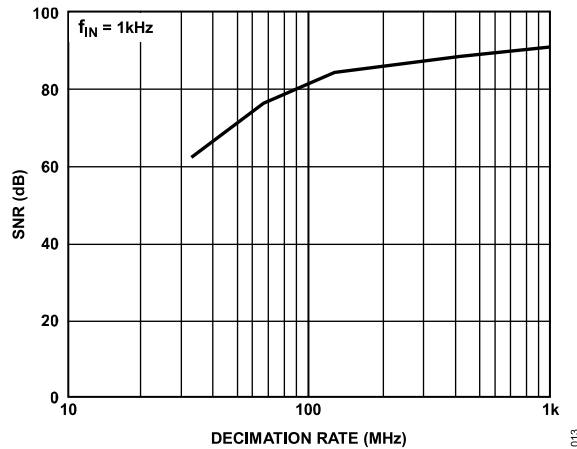


Figure 30. SNR vs. Decimation Rate of Sinc3 Filter Order

Table 13. Sinc3 Filter Characteristics for 20 MHz MCLKIN

Decimation Ratio (DR)	Throughput Rate (kHz)	Output Data Size (Bits)	Filter Response (kHz)
32	625	15	163.7
64	312.5	18	81.8
128	156.2	21	40.9
256	78.1	24	20.4
512	39.1	27	10.2

A sinc3 filter is recommended for the ADuM7703. This filter can be implemented on a field programmable gate array (FPGA) or a digital signal processor (DSP). Equation 9 describes the transfer function of a sinc filter.

$$H(Z) = \left(\frac{1}{DR} \frac{(1 - Z^{-DR})}{(1 - Z^{-1})} \right)^N \quad (9)$$

where:
 Z is the sample.
 DR is the decimation rate.
 N is the sinc filter order.

The throughput rate of the sinc filter is determined by the modulator clock and the decimation rate selected.

$$\text{Throughput} = \text{MCLK}/DR \quad (10)$$

where $MCLK$ is the modulator clock frequency.

As the decimation rate increases, the data output size from the sinc filter increases. The output data size is expressed in Equation 11. The 16 most significant bits are used to return a 16-bit result.

$$\text{Data Size} = N \times \log_2 DR \quad (11)$$

For a sinc3 filter, the -3 dB filter response point can be derived from the filter transfer function, Equation 9, and is 0.262 times the throughput rate. The filter characteristics for a third-order sinc filter are summarized in Table 13.

APPLICATIONS INFORMATION

INTERFACING TO ADSP-CM4XX

The **ADSP-CM4xx** family of mixed-signal control processors contains an on-chip sinc filter and clock generation modules for direct connection to the ADuM7703 MCLKIN and MDAT pins. The ADSP-CM4xx can process bit streams from four ADuM7703 devices using a pair of configurable sinc filters for each bit stream. The primary sinc filter of each pair produces the filtered and decimated output for the pair. The output can be decimated to any integer rate between 8 times and 256 times lower than the input rate. The four secondary sinc filters are low latency filters with programmable positive and negative overrange detection comparators that can detect system fault conditions.

Figure 31 shows the typical interface between the ADuM7703 and the ADSP-CM4xx. Additional information on the configuration of the sinc filter modules in the ADSP-CM4xx can be found in the [AN-1265 Application Note](#).

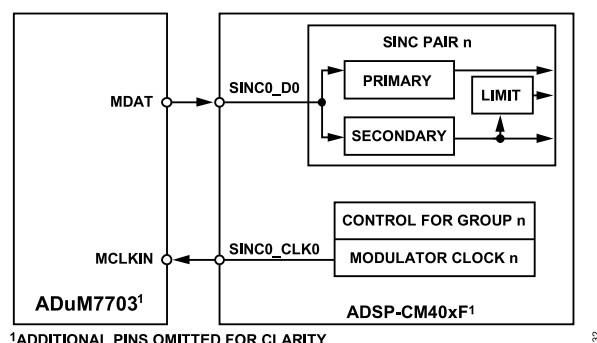


Figure 31. Interfacing the ADuM7703 to the ADSP-CM4xx

GROUNDING AND LAYOUT

It is recommended to decouple the V_{DD1} supply with a 10 μF capacitor in parallel with a 100 nF capacitor to GND_1 . Decouple the V_{DD2} supply with a 10 μF capacitor in parallel with a 100 nF capacitor to GND_2 . In applications involving high common-mode transients, ensure that board coupling across the isolation barrier is minimized. Furthermore, design the board layout so that any coupling that occurs equally affects all pins on a given component side. Failure to ensure equal coupling can cause voltage differentials between pins to exceed the absolute maximum ratings of the device, thereby leading to latch-up or permanent damage. Place any decoupling as close to the supply pins as possible.

Minimize series resistance in the analog inputs to avoid any distortion effects, especially at high temperatures. If possible, equalize the source impedance on each analog input to minimize offset. Check for mismatch and thermocouple effects on the analog input PCB tracks to reduce offset drift.

INSULATION LIFETIME

All insulation structures eventually break down when subjected to voltage stress over a sufficiently long period. The rate of insulation degradation is dependent on the characteristics of the voltage waveform applied across the insulation. In addition to the testing performed by the regulatory agencies, Analog Devices carries out an extensive set of evaluations to determine the lifetime of the insulation structure within the ADuM7703.

Analog Devices performs accelerated life testing using voltage levels higher than the rated continuous working voltage. Acceleration factors for several operating conditions are determined. These factors allow calculation of the time to failure at the actual working voltage. The values shown in [Table 9](#) summarize the peak voltage for 37.5 years of (reinforced) service life for a bipolar, ac operating condition and the maximum VDE approved working voltages.

These tests subjected the ADuM7703 to continuous cross isolation voltages. To accelerate the occurrence of failures, the selected test voltages were values exceeding those of normal use. The time to failure values of these units were recorded and used to calculate the acceleration factors. These factors were then used to calculate the time to failure under the normal operating conditions. The values shown in [Table 9](#) are the lesser of the following two values:

- The value that ensures at least a 37.5 year lifetime of continuous (reinforced) use.
- The maximum VDE approved working voltage.

The lifetime of the ADuM7703 is guaranteed using a bipolar ac waveform, as shown in [Figure 32](#).

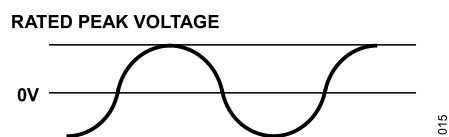


Figure 32. Bipolar AC Waveform, 50 Hz or 60 Hz

OUTLINE DIMENSIONS

Package Drawing (Option)	Package Type	Package Description
RW-16	SOIC_W	16-Lead Standard Small Outline Package, Wide Body
RI-8-1	SOIC_IC	8-Lead Standard Small Outline Package with Increased Creepage, Wide Body

For the latest package outline information and land patterns (footprints), go to [Package Index](#).

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Packing Quantity	Package Option
ADuM7703BRWZ	-40°C to +125°C	16-Lead Standard Small Outline Package [SOIC_W]	Tube, 47	RW-16
ADuM7703BRWZ-RL	-40°C to +125°C	16-Lead Standard Small Outline Package [SOIC_W]	Reel, 1000	RW-16
ADuM7703BRWZ-RL7	-40°C to +125°C	16-Lead Standard Small Outline Package [SOIC_W]	Reel7, 400	RW-16
ADuM7703-8BRIZ	-40°C to +105°C	8-Lead Standard Small Outline Package, with Increased Creepage [SOIC_IC]	Tube, 80	RI-8-1
ADuM7703-8BRIZ-RL	-40°C to +105°C	8-Lead Standard Small Outline Package, with Increased Creepage [SOIC_IC]	Reel, 1500	RI-8-1
ADuM7703-8BRIZ-RL7	-40°C to +105°C	8-Lead Standard Small Outline Package, with Increased Creepage [SOIC_IC]	Reel7, 400	RI-8-1

¹ Z = RoHS-Compliant Part.

EVALUATION BOARDS

Model ^{1, 2}	Description
EV-ADuM7703-8FMCZ	Evaluation Board

¹ Z = RoHS-Compliant Part.

² The EV-ADuM7703-8FMCZ is compatible with the [EVAL-SDP-CH1Z](#) high speed controller board.