

Flip Chip, Silicon SP4T Switch, 1GHz to 81GHz

FEATURES

- ▶ Frequency range: 1GHz to 81GHz
- ▶ Low insertion loss
 - ▶ 1.9dB typical to 40GHz
 - ▶ 3.0dB typical to 67GHz
 - ▶ 4.0dB typical to 81GHz
- ▶ High Isolation
 - ▶ 39dB typical up to 40GHz
 - ▶ 30dB typical up to 67GHz
 - ▶ 20dB typical up to 81GHz
- ▶ High input linearity
 - ▶ P0.1dB: 25dBm typical
 - ▶ IP3: 47dBm typical
- ▶ High RF input power handling
 - ▶ Through path: 24dBm
 - ▶ Hot switching: 24dBm
- ▶ No low frequency spurs
- ▶ CMOS-/LVTTTL-compatible
- ▶ Fast on time and off time: 25ns
- ▶ RF settling time (50% V_{CTRL} to 0.1dB of final RF output): 35ns
- ▶ Single-supply operation capability ($VDD = 3.3V$ and $VSS = 0V$)
- ▶ [56-ball, 2.390mm × 2.540mm, bumped, bare die sales](#)

APPLICATIONS

- ▶ Test and instrumentation
- ▶ Cellular infrastructure: 5G mmWave
- ▶ Military radios, radars, electronic counter measures (ECMs)
- ▶ Microwave radios and very small aperture terminals (VSATs)

FUNCTIONAL BLOCK DIAGRAM

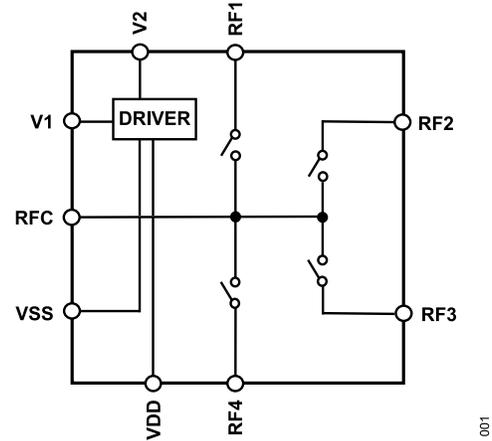


Figure 1. Functional Block Diagram

GENERAL DESCRIPTION

The ADRF5440 is a reflective, SP4T switch manufactured in the silicon process. This switch operates from 1GHz to 81GHz with better than 4.0dB of insertion loss and 20dB of isolation. The ADRF5440 has an RF input power handling capability of 24dBm for the through path and 24dBm for hot switching.

The ADRF5440 draws a current of 145 μ A on the positive supply of +3.3V and 510 μ A on negative supply of -3.3V. The device employs complementary metal-oxide semiconductor (CMOS)-/low voltage transistor to transistor logic (LVTTTL)-compatible controls.

The ADRF5440 can also operate with a single positive supply voltage (VDD) applied while the negative supply voltage (VSS) is tied to ground. The small signal performance is maintained while the switching characteristics, linearity, and power handling performance is derated. See [Table 2](#) for more details.

The ADRF5440 RF ports are designed to match a characteristic impedance of 50 Ω . The ADRF5440 is [56-ball, 2.390mm × 2.540mm, bumped bare die sales](#) and can operate from -40°C to +105°C.

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REVISION HISTORY**11/2025—Rev. 0 to Rev. A**

Change to Features Section.....	1
Changes to Applications Section.....	1
Changes to General Description Section.....	1
Changes to Table 6.....	7
Change to Insertion Loss, Return Loss, and Isolation Section.....	10
Changes to Input Power Compression and Third-Order Intercept Section.....	12
Changes to Recommendations for PCB Design Section, Figure 26, Figure 27 Caption, and Table 8 Title...	16
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Changes to Reflow Assembly with Flux Dipping Section.....	17
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9/2025—Revision 0: Initial Version

SPECIFICATIONS

ELECTRICAL SPECIFICATIONS

VDD = 3.3V, VSS = -3.3V, V1 = V2 = 0V or VDD, and T_{CASE} = 25°C for 50Ω system, unless otherwise noted. RFx refers to RF1, RF2, RF3, and RF4.

Table 1. Electrical Specifications

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
FREQUENCY RANGE	f		1		81	GHz
INSERTION LOSS						
Between RFC and RFx (On)		100MHz to 18GHz		1.4		dB
		18GHz to 40GHz		1.9		dB
		40GHz to 55GHz		2.4		dB
		55GHz to 67GHz		3.0		dB
		67GHz to 81GHz		4.0		dB
RETURN LOSS						
RFC and RFx (On)		100MHz to 18GHz		17		dB
		18GHz to 40GHz		13		dB
		40GHz to 55GHz		13		dB
		55GHz to 67GHz		15		dB
		67GHz to 81GHz		12		dB
ISOLATION						
Between RFC and RFx (Off)		100MHz to 18GHz		45		dB
		18GHz to 40GHz		39		dB
		40GHz to 55GHz		35		dB
		55GHz to 67GHz		30		dB
		67GHz to 81GHz		20		dB
Between RFx and RFx		100MHz to 18GHz		43		dB
		18GHz to 40GHz		38		dB
		40GHz to 55GHz		35		dB
		55GHz to 67GHz		30		dB
		67GHz to 81GHz		15		dB
SWITCHING CHARACTERISTICS						
Rise Time and Fall Time	t _{RISE} , t _{FALL}	10% to 90% of RF output (RF _{OUT})		5		ns
On Time and Off Time	t _{ON} , t _{OFF}	50% V _{CTRL} to 90% of RF _{OUT}		25		ns
RF Settling Time		50% V _{CTRL} to 0.1dB of final RF _{OUT}		35		ns
0.1dB						
INPUT LINEARITY ¹		1GHz to 67GHz				
0.1dB Power Compression	P0.1dB			25		dBm
Third-Order Intercept	IP3	Two-tone input power = 14dBm each tone, Δf = 1MHz		47		dBm
SUPPLY CURRENT		VDD and VSS pins				
Positive Supply Current	I _{DD}			145		μA
Negative Supply Current	I _{SS}			510		μA
DIGITAL CONTROL INPUTS		V1 and V2 pins				
Voltage						
Low	V _{INL}		0		0.8	V
High	V _{INH}		1.2		3.3	V
Current						
Low and High	I _{INL} , I _{INH}			<1		μA

SPECIFICATIONS

Table 1. Electrical Specifications (Continued)

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
RECOMMENDED OPERATING CONDITIONS						
Supply Voltage						
Positive	V _{DD}		3.15		3.45	V
Negative	V _{SS}		-3.45		-3.15	V
Digital Control Voltage	V _{CTRL}		0		V _{DD}	V
RF Input Power ²	P _{IN}	f = 3GHz to 70GHz, T _{CASE} = 85°C ³				
Through Path		RF signal is applied to RFC or through a connected RFx			24	dBm
Hot Switching		RF signal is present at RFC while switching between RFx			24	dBm
Case Temperature	T _{CASE}		-40		+105	°C

¹ For input linearity performance over frequency, see Figure 21 to Figure 24.

² For power derating over frequency, see Figure 2 and Figure 3.

³ For 105°C operation, the power handling degrades from the T_{CASE} = 85°C specification by 3dB.

SINGLE-SUPPLY OPERATION

V_{DD} = 3.3V, V_{SS} = 0V, V1 = 0V or V_{DD}, V2 = 0V or V_{DD}, T_{CASE} = 25°C for 50Ω system, unless otherwise noted.

Table 2. Single-Supply Operation Specifications

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
FREQUENCY RANGE	f		1		81	GHz
SWITCHING CHARACTERISTICS						
Rise Time and Fall Time	t _{RISE} , t _{FALL}	10% to 90% of RF _{OUT}		20		ns
On Time and Off Time	t _{ON} , t _{OFF}	50% V _{CTRL} to 90% of RF _{OUT}		58		ns
0.1dB RF Settling Time		50% V _{CTRL} to 0.1dB of final RF _{OUT}		62		ns
INPUT LINEARITY						
0.1dB Power Compression	P0.1dB	f = 1GHz to 81GHz		13		dBm
Input Third-Order Intercept	IIP3	Two-tone input power = 0dBm each tone, Δf = 1MHz		41		dBm
RECOMMENDED OPERATING CONDITIONS						
RF Input Power ¹	P _{IN}	f = 3GHz to 70GHz, T _{CASE} = 85°C				
Through Path		RF signal is applied to the RFC or through connected RFx			13	dBm
Hot Switching		RF signal is applied to the RFC while switching between RFx			13	dBm

¹ For 105°C operation, the power handling degrades from the T_{CASE} = 85°C specification by 1dB.

ABSOLUTE MAXIMUM RATINGS

For the recommended operating conditions, see [Table 1](#).

Table 3. Absolute Maximum Ratings

Parameter	Rating
Positive Supply Voltage	-0.3V to +3.6V
Negative Supply Voltage	-3.6V to +0.3V
Digital Control Input Voltage ¹	
Voltage	-0.3V to VDD + 0.3V
Current	3mA
RF Input Power, Dual Supply ² (VDD = 3.3V, VSS = -3.3V, f = 3GHz to 70GHz, T _{CASE} = 85°C ³)	
Through Path	25dBm
Hot Switching (RFC)	25dBm
RF Input Power, Single Supply (VDD = 3.3V, VSS = 0V, f = 3GHz to 70GHz, T _{CASE} = 85°C ³)	
Through Path	14dBm
Hot Switching (RFC)	14dBm
RF Input Power, Unbiased (VDD and VSS = 0V)	14dBm
Temperature	
Junction, T _J	135°C
Storage Range	-65°C to +150°C
Reflow	260°C

¹ Overvoltages at digital control inputs are clamped by internal diodes. Current must be limited to the maximum rating given.

² For power derating over frequency, see [Figure 2](#) and [Figure 3](#).

³ For 105°C operation, the power handling degrades from the T_{CASE} = 85°C specification by 3dB for dual supply and 1dB for single supply.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

Only one absolute maximum rating can be applied at any one time.

THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

θ_{JC} is the junction to case bottom (channel to package bottom) thermal resistance.

Table 4. Thermal Resistance

Package Type	θ_{JC} ¹	Unit
CD-56-1, Through Path	360	°C/W

¹ θ_{JC} was determined by simulation under the following conditions: the heat transfer is due solely to the thermal conduction from the channel through the ground pad to the PCB, and the ground pad is held constant at the operating temperature of 85°C.

POWER DERATING CURVES

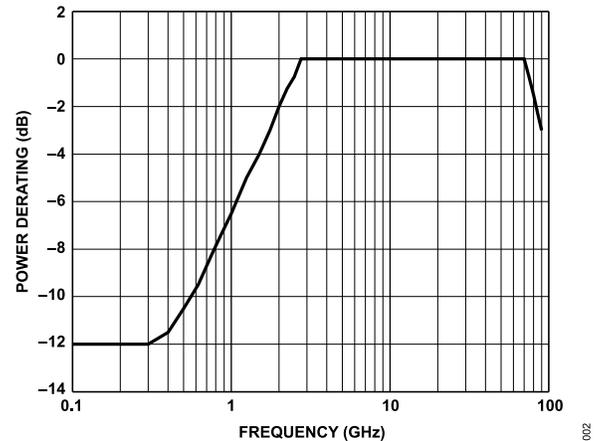


Figure 2. Power Derating vs. Frequency, Low Frequency Detail, T_{CASE} = 85°C

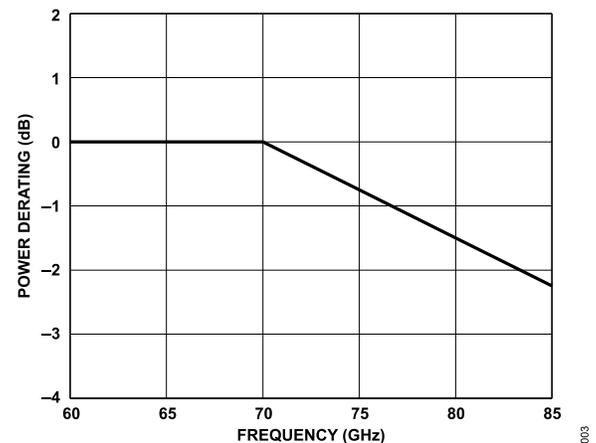


Figure 3. Power Derating vs. Frequency, High Frequency Detail, T_{CASE} = 85°C

ABSOLUTE MAXIMUM RATINGS**ELECTROSTATIC DISCHARGE (ESD) RATINGS**

The following ESD information is provided for handling of ESD-sensitive devices in an ESD protected area only.

Human body model (HBM) per ANSI/ESDA/JEDEC JS-001.

ESD Ratings for the ADRF5440

Table 5. ADRF5440, 56-Ball BUMPED_CHIP

ESD Model	Withstand Threshold (V)
HBM	±1000 for RFx pins ±2000 for supply and digital control pins

ESD CAUTION

ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

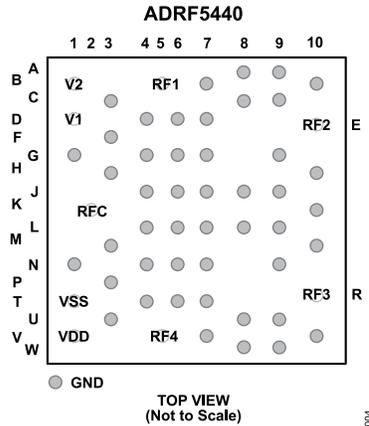


Figure 4. Pin Configuration (Top View—Ball Side Down)

Table 6. Pin Function Descriptions

Ball No.	Mnemonic	X Coordinate (mm)	Y Coordinate (mm)	Description
A8	GND	+0.400	+1.150	Ground.
A9	GND	+0.700	+1.150	Ground.
B1	V2	-1.0	+1.050	Control Input Voltage. See Figure 5 for the interface schematic.
B5	RF1	-0.275	+1.050	RF Port 1. The RF1 pin is DC-coupled and biased to 0V and AC matched to 50Ω. No DC blocking capacitor is necessary when the RF line potential is equal to 0V DC. See Figure 6 for the interface schematic.
B7	GND	+0.100	+1.050	Ground.
B10	GND	+1.000	+1.050	Ground.
C3	GND	-0.700	+0.900	Ground.
C8	GND	+0.400	+0.900	Ground.
C9	GND	+0.700	+0.900	Ground.
D1	V1	-1.0	+0.750	Control Input Voltage. See Figure 5 for the interface schematic.
D4	GND	-0.400	+0.750	Ground.
D6	GND	-0.150	+0.750	Ground.
D7	GND	+0.100	+0.750	Ground.
E10	RF2	+1.000	+0.700	RF Port 2. The RF2 pin is DC-coupled and biased to 0V and AC matched to 50Ω. No DC blocking capacitor is necessary when the RF line potential is equal to 0V DC. See Figure 6 for the interface schematic.
F3	GND	-0.700	+0.600	Ground.
G1	GND	-1.0	+0.450	Ground.
G4	GND	-0.400	+0.450	Ground.
G6	GND	-0.150	+0.450	Ground.
G7	GND	+0.100	+0.450	Ground.
G9	GND	+0.700	+0.450	Ground.
H3	GND	-0.700	+0.300	Ground.
H10	GND	+1.000	+0.300	Ground.
J4	GND	-0.400	+0.150	Ground.
J6	GND	-0.150	+0.150	Ground.
J7	GND	+0.100	+0.150	Ground.
J8	GND	+0.400	+0.150	Ground.
J9	GND	+0.700	+0.150	Ground.
K2	RFC	-0.850	+0.000	RF Common Port. The RFC pin is DC-coupled and biased to 0V and AC matched to 50Ω. No DC blocking capacitor is necessary when the RF line potential is equal to 0V DC. See Figure 6 for the interface schematic.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

Table 6. Pin Function Descriptions (Continued)

Ball No.	Mnemonic	X Coordinate (mm)	Y Coordinate (mm)	Description
K10	GND	+1.000	+0.000	Ground.
L4	GND	-0.400	-0.150	Ground.
L6	GND	-0.150	-0.150	Ground.
L7	GND	+0.100	-0.150	Ground.
L8	GND	+0.400	-0.150	Ground.
L9	GND	+0.700	-0.150	Ground.
M3	GND	-0.700	-0.300	Ground.
M10	GND	+1.000	-0.300	Ground.
N1	GND	-1.0	-0.450	Ground.
N4	GND	-0.400	-0.450	Ground.
N6	GND	-0.150	-0.450	Ground.
N7	GND	+0.100	-0.450	Ground.
N9	GND	+0.700	-0.450	Ground.
P3	GND	-0.700	-0.600	Ground.
R10	RF3	+1.000	-0.700	RF Port 3. The RF3 pin is DC-coupled and biased to 0V and AC matched to 50Ω. No DC blocking capacitor is necessary when the RF line potential is equal to 0V DC. See Figure 6 for the interface schematic.
T1	VSS	-1.0	-0.750	Negative Supply Voltage. See Figure 7 for the interface schematic.
T4	GND	-0.400	-0.750	Ground.
T6	GND	-0.150	-0.750	Ground.
T7	GND	+0.100	-0.750	Ground.
U3	GND	-0.700	-0.900	Ground.
U8	GND	+0.400	-0.900	Ground.
U9	GND	+0.700	-0.900	Ground.
V1	VDD	-1.0	-1.050	Positive Supply Voltage. See Figure 8 for the interface schematic.
V5	RF4	-0.275	-1.050	RF Port 4. The RF4 pin is DC-coupled and biased to 0V and AC matched to 50Ω. No DC blocking capacitor is necessary when the RF line potential is equal to 0V DC. See Figure 6 for the interface schematic.
V7	GND	+0.100	-0.150	Ground.
V10	GND	+1.000	-1.050	Ground.
W8	GND	+0.400	-1.150	Ground.
W9	GND	+0.700	-1.150	Ground.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS

INTERFACE SCHEMATICS

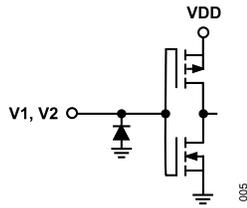


Figure 5. V1, V2 Interface Schematic

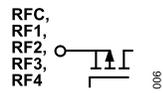


Figure 6. RFC, RF1, RF2, RF3, and RF4 Interface Schematic

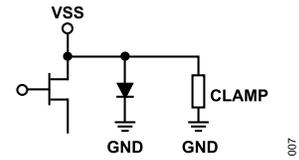


Figure 7. VSS Pin Interface Schematic

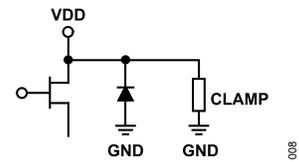


Figure 8. VDD Pin Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS

INSERTION LOSS, RETURN LOSS, AND ISOLATION

VDD = 3.3V, VSS = -3.3V, V1 = V2 = 0V or VDD, and T_{CASE} = 25°C for a 50Ω system, unless otherwise noted. Insertion loss, return loss, and isolation are measured on the probe board using ground-signal-ground (GSG) probes close to the RFx pins.

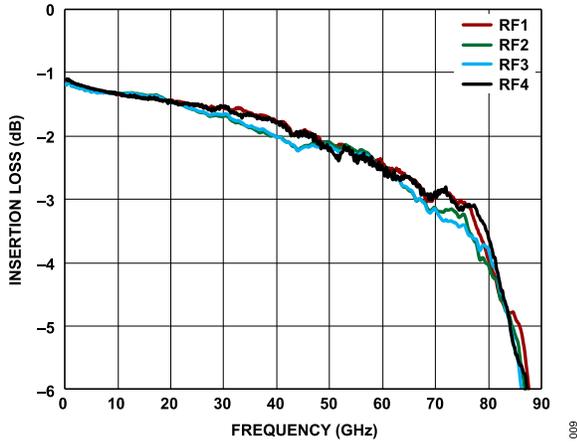


Figure 9. Insertion Loss for RFC to RFx On vs. Frequency

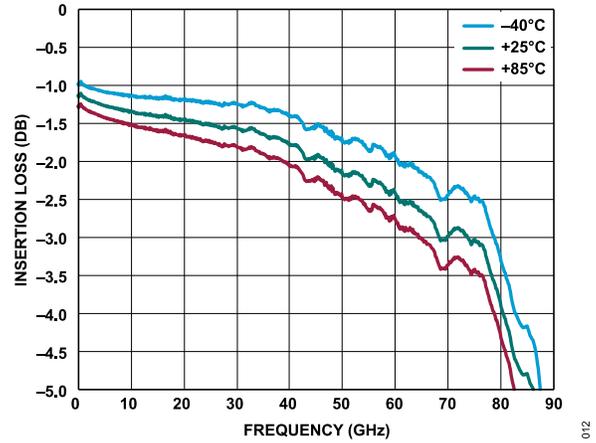


Figure 12. Insertion Loss for RFC to RF1 On vs. Frequency over Various Temperature

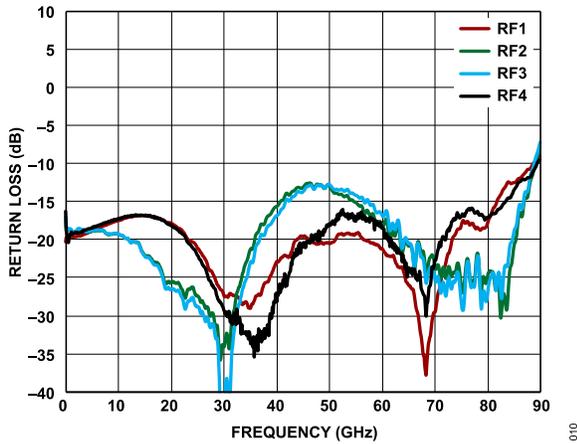


Figure 10. Return Loss for RFC when RFx Selected vs. Frequency

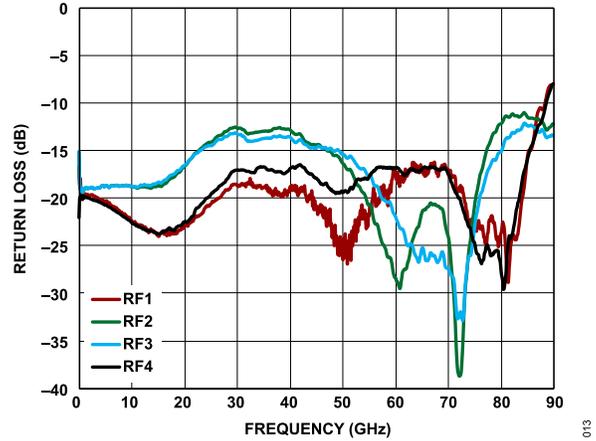


Figure 13. Return Loss for RFx Selected vs. Frequency

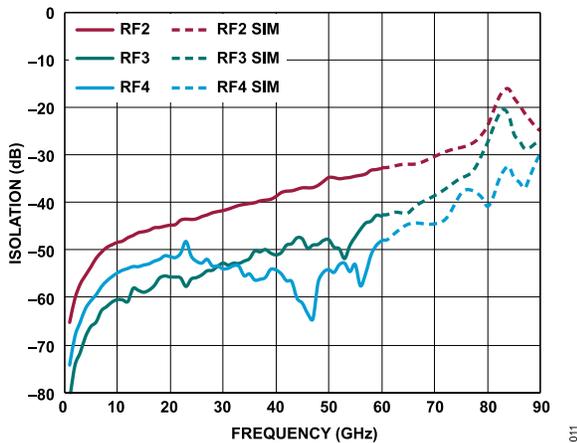


Figure 11. Isolation for RFC to RFx Off vs. Frequency, RFC to RF1 Path On (SIM Is Simulated Data)

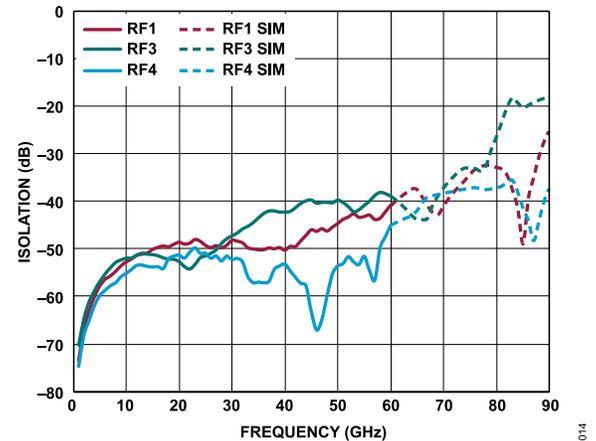


Figure 14. Isolation for RFC to RFx Off vs. Frequency, RFC to RF2 Path On

TYPICAL PERFORMANCE CHARACTERISTICS

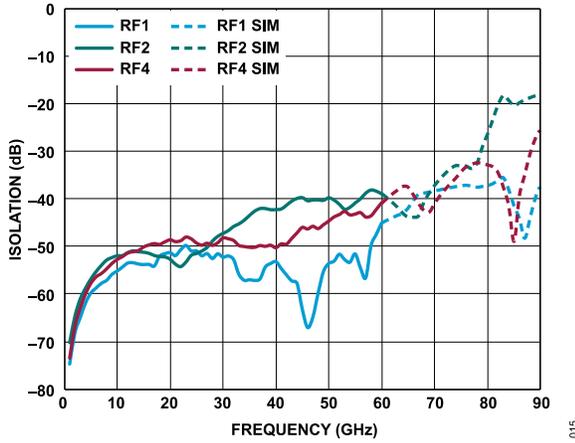


Figure 15. Isolation for RFC to RFx Off vs. Frequency, RFC to RF3 Path On

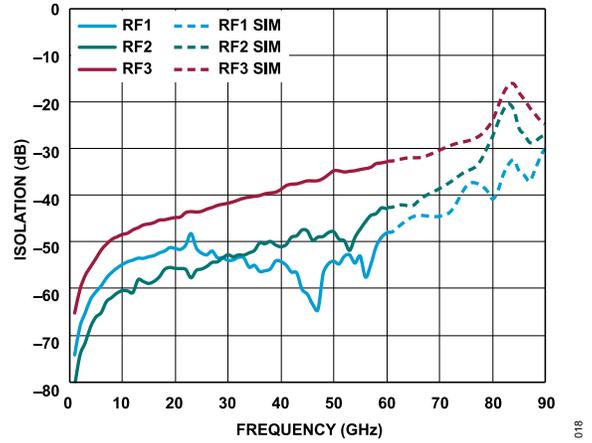


Figure 18. Isolation for RFC to RFx Off vs. Frequency, RFC to RF4 Path On

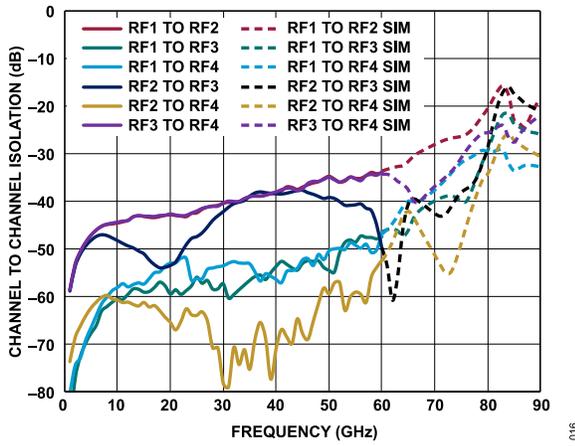


Figure 16. Channel to Channel Isolation vs. Frequency, RFC to RF1 Path On

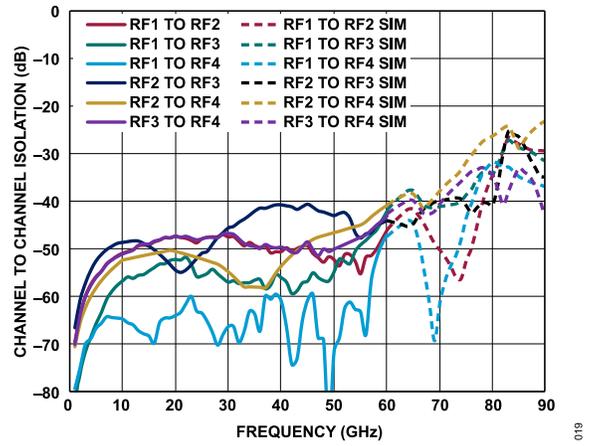


Figure 19. Channel to Channel Isolation vs. Frequency, RFC to RF2 Path On

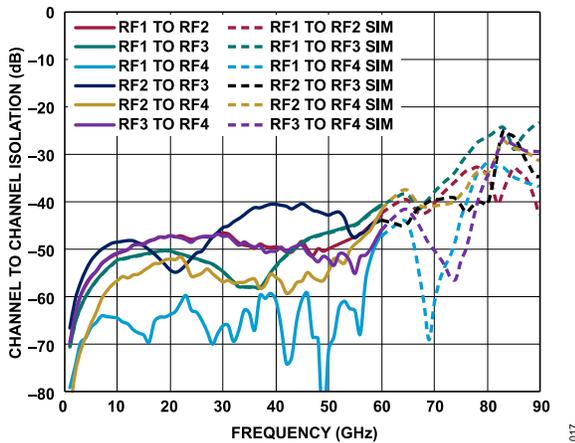


Figure 17. Channel to Channel Isolation vs. Frequency, RFC to RF3 Path On

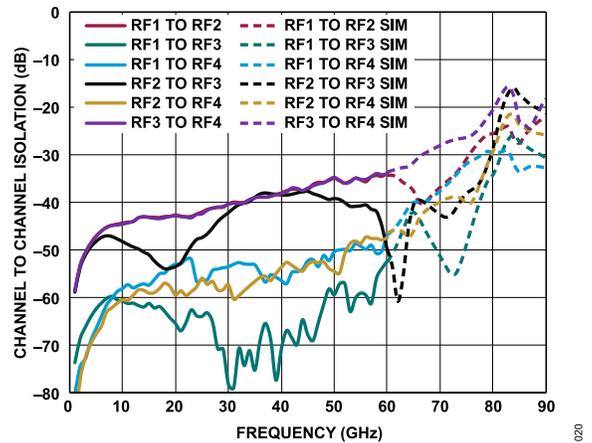


Figure 20. Channel to Channel Isolation vs. Frequency, RFC to RF4 Path On

TYPICAL PERFORMANCE CHARACTERISTICS

INPUT POWER COMPRESSION AND THIRD-ORDER INTERCEPT

VDD = 3.3V, VSS = -3.3V, V1 = V2 = 0V or VDD, and T_{CASE} = 25°C for a 50Ω system, unless otherwise noted. The large signal performance parameters are measured on a connectorized board.

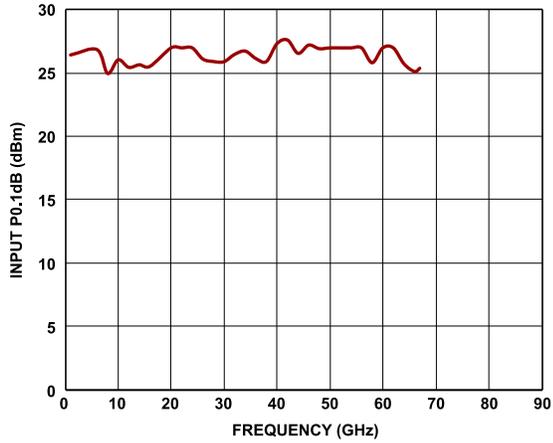


Figure 21. Input P0.1dB vs. Frequency

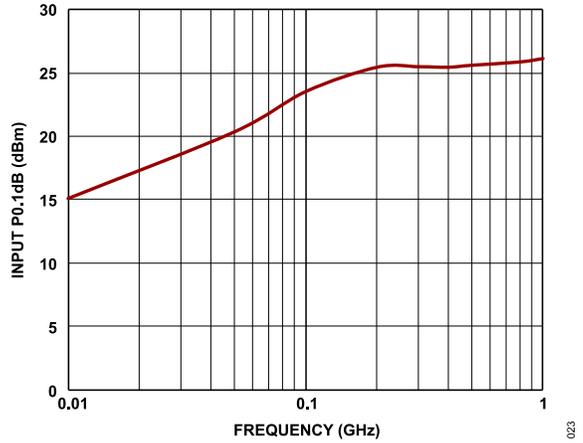


Figure 23. Input P1dB vs. Frequency (Low Frequency Detail)

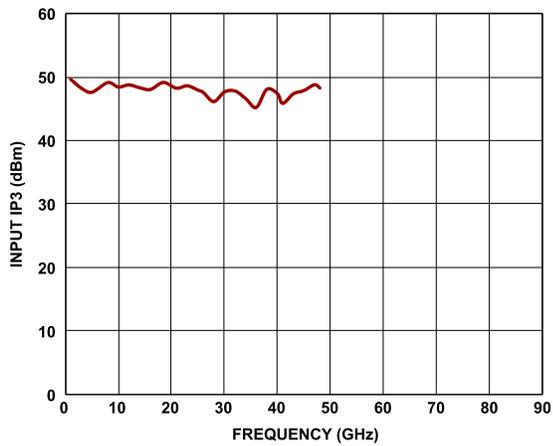


Figure 22. Input IP3 vs. Frequency

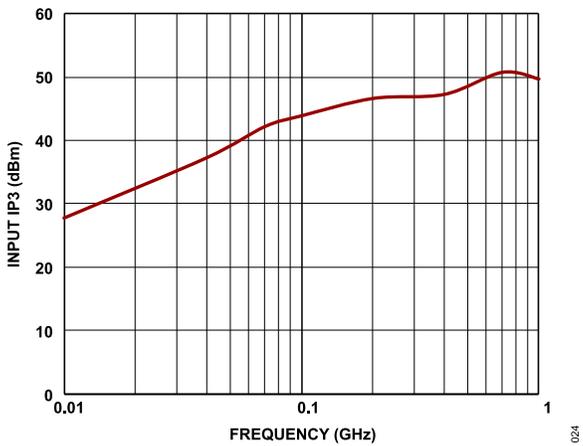


Figure 24. Input IP3 vs. Frequency (Low Frequency Detail)

THEORY OF OPERATION

The ADRF5440 integrates a driver to perform logic functions internally and provides the user with the advantage of a simplified CMOS-/LVTTTL-compatible control interface. This driver features two

digital control input pins, V1 and V2. The logic level applied to the control pins determine which RF port is in the insertion loss state and in the isolation state (see [Table 7](#)).

Table 7. Control Voltage Truth Table

Digital Control Inputs		RFx Paths			
V1	V2	RFC to RF1	RFC to RF2	RFC to RF3	RFC to RF4
Low	Low	Insertion loss (on)	Isolation (off)	Isolation (off)	Isolation (off)
High	Low	Isolation (off)	Insertion loss (on)	Isolation (off)	Isolation (off)
Low	High	Isolation (off)	Isolation (off)	Insertion loss (on)	Isolation (off)
High	High	Isolation (off)	Isolation (off)	Isolation (off)	Insertion loss (on)

THEORY OF OPERATION

RF INPUT AND OUTPUT

The RF ports (RFC, RF1 to RF4) are DC-coupled to 0V, and no DC blocking is required at the RF ports when the RF line potential is equal to 0V. The RF ports are internally matched to 50Ω.

The ADRF5440 is bidirectional with equal power handling capabilities. The RF input signal can be applied to the RFC port or the selected RF throw port.

The insertion loss path conducts the RF signal between the selected RF throw port and the RF common port. The isolation paths provide high loss between the insertion loss path and the unselected RF throw ports. The unselected RF ports of the ADRF5440 are reflective.

The power handling of the ADRF5440 derates with frequency less than 3GHz and more than 70GHz. See [Figure 2](#) and [Figure 3](#) for the derating of the RF power towards lower and higher frequencies.

POWER SUPPLY

The ADRF5440 requires a positive supply voltage applied to the VDD pin and a negative supply voltage applied to the VSS pin. Bypassing capacitors are recommended on the supply lines to filter high frequency noise.

The ideal power-up sequence is as follows:

1. Connect to GND.
2. Power up VDD and VSS. Power up VSS after VDD to avoid current transients on VDD during ramp up.
3. Apply the digital control inputs (V1 and V2). The relative order of the control inputs is not important. However, powering the digital control inputs before the VDD supply can inadvertently forward bias and damage the internal ESD protection structures. To avoid this damage, use a series 1kΩ resistor to limit the current flowing into the control pins. Use pull-up or pull-down resistors if the controller is in a high impedance state after VDD is powered up and the control pins are not driven to a valid logic state.
4. Apply an RF input signal.

The ideal power-down sequence is the reverse order of the power-up sequence.

APPLICATIONS INFORMATION

The ADRF5440 has two power supply pins (VDD and VSS) and two digital control pins (V1 and V2). [Figure 25](#) shows the external components and connections for the supply pins. The VDD, VSS, V1, and V2 pins are decoupled with a 100pF multilayer ceramic capacitor. The device pinout allows the placement of the decoupling capacitors close to the device. No other external components are needed for bias and operation, except DC blocking capacitors on the RFX pins when the RF lines are biased at a voltage different than 0V. Refer to the [Pin Configuration and Function Descriptions](#) section for further details.

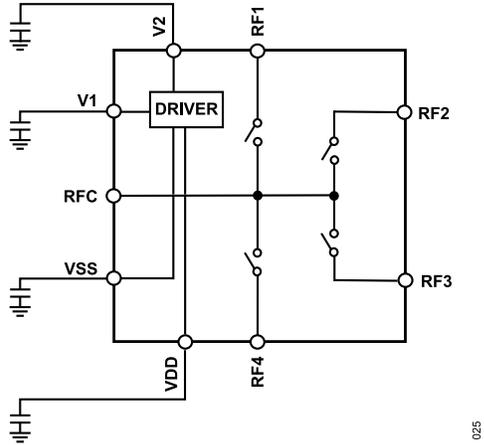


Figure 25. Recommended Schematic

APPLICATIONS INFORMATION

RECOMMENDATIONS FOR PCB DESIGN

The RF ports are matched to 50Ω internally, and the pinout is designed to mate a coplanar waveguide (CPWG) with 50Ω characteristic impedance on the PCB. Figure 26 shows the referenced CPWG design for a 150μm thick Megtron6 2×1080 R-5775G dielectric material. The RF trace with 300μm width and 300μm gap is used for 42μm finished copper thickness.

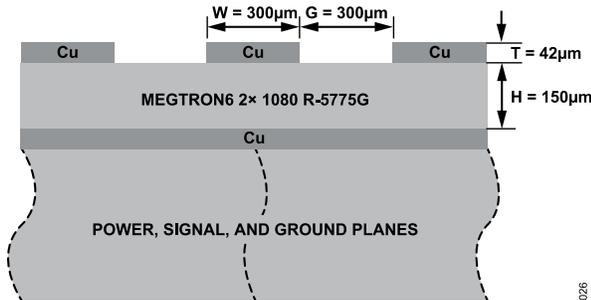


Figure 26. Probe Board Stackup

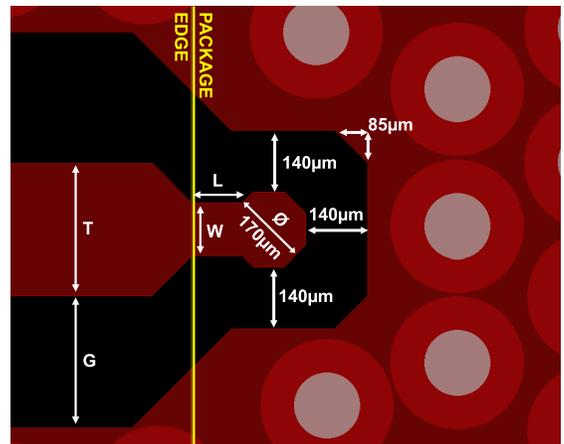


Figure 27. Recommended RF Pin Transitions

Table 8 shows the recommended layout from the device RF pins to the 50Ω CPWG. Signal pads have a 170μm diameter extended with a neck for broadband tuning and tapered to RF trace with a 45° angle. The optimized RF pin transitions for broadband performance is given for different stackups in Table 8. For further recommendations on alternate PCB stackups with different dielectric thickness and CPWG design, contact Analog Devices, Inc., Analog Devices Support.

Table 8. Recommended RF Pin Transitions for Different Stackups

RF Stack-Up Height and Dielectric		RF Trace Dimensions		ADRF5440 Mnemonic	Transition Dimensions	
H (μm)	Er	W (μm)	G (μm)		N (μm)	L (μm)
85	3.02	180	250	RFC, RF1, and RF4 RF2 and RF3	120	90
100	3.34	200	225	RFC, RF1, and RF4 RF2 and RF3	120	90
115	3.40	225	250	RFC, RF1, and RF4 RF2 and RF3	120	140
125	3.00	250	225	All	120	90
150 ¹	3.40	300	300	All	120	90
175	3.16	360	245	All	170	90
190	3.00	425	225	All	170	90
200	3.55	425	275	All	170	90

¹ Design reference and generic footprint.

APPLICATIONS INFORMATION

DIE ASSEMBLY

The ADRF5440 complies with the standard RoHS reflow assembly process and can be assembled together with other surface-mounted technology (SMT) components in the same reflow cycle. The PCB must incorporate fiducials close to the device to get the best pick and place accuracy.

The top copper layer of the PCB is designed for optimum RF performance, and the solder mask and paste mask layers are designed for optimum assembly yield. The ground pads are drawn as solder mask defined. The signal pads are drawn as pad defined. The same solder mask and paste mask design is used for both pads. Alternatively, the ADRF5440 can be assembled without applying a solder paste on the PCB. If no solder paste is applied, the device must be dipped into flux prior to placement on the PCB.

Reflow Assembly with Solder Paste

Solder mask openings of $175\mu\text{m}$ in a square shape are recommended for the signal and RF pads, and solder mask openings of $150\mu\text{m}$ in a square shape are recommended for the GND pads. Solder mask thickness must not exceed $50\mu\text{m}$. Paste mask is drawn circular with a $150\mu\text{m}$ diameter. Using a stencil with 2mil thickness and no aperture, reduction yields the optimum paste mask print. The device does not need any flux dipping during the pick and place process.

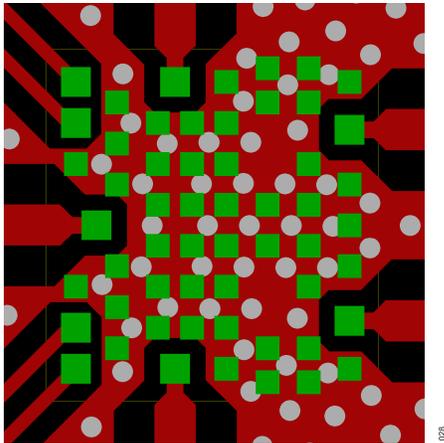


Figure 28. Recommended Footprint for Solder Paste

Reflow Assembly with Flux Dipping

Solder mask openings in a circular shape and $130\mu\text{m}$ in diameter are recommended. Solder mask thickness must not exceed $50\mu\text{m}$. Solder paste is not applied. The device is dipped into flux prior to placement on the board.

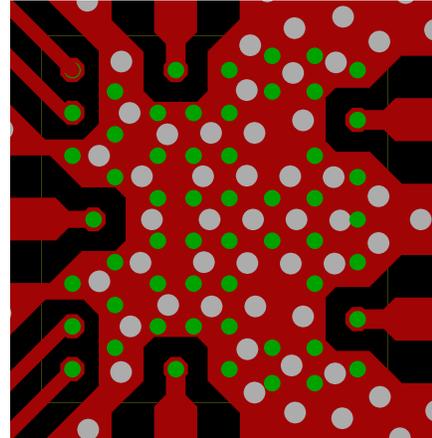


Figure 29. Recommended Footprint for Flux Dip Assembly

Assembly Durability

The device is a bumped die. The bump composition is copper (Cu) pillar with a plated tin (Sn) and silver (Ag) solder-cap, which forms the device interconnect to PCB when it is attached with reflow. The integrity of the interconnect structures and solder joints are compromised by the accumulated stress over temperature due to a mismatch of the thermomechanical properties of the different materials in the complete assembly of the user hardware. The ADRF5440 fulfills the JEDEC JED47L requirements under the JESD22-A104 condition J (0-100C) for 2300 temperature cycles while mounted on a 0.75mm thick laminate PCB, which is assembled using solder paste reflow and without any underfill material applied. Users must qualify their hardware for the environmental conditions aligned with their end-use requirements because differences in PCB properties and design, varied number of temperature cycles, soak and dwell times, changes in temperature range, can affect thermomechanical results. Note that using a proper underfill adhesive significantly improves the integrity of the device assembly by reducing the stress exposed to the interconnect structures and solder joints.

OUTLINE DIMENSIONS

Package Drawing Option	Package Type	Package Description
CD-56-1	BUMPED_CHIP	56-Ball Bumped Bare Die Sales [BUMPED_CHIP]

For the latest package outline information and land patterns (footprints), go to [Package Index](#).

ORDERING GUIDE

Model ¹	Temperature Range	Package Description	Packaging Quantity	Package Option
ADRF5440BCDZ	-40°C to +105°C	56-Ball Bumped Bare Die Sales [BUMPED_CHIP]	Cut-Tape, 1 to 500	CD-56-1
ADRF5440BCDZ-R7	-40°C to +105°C	56-Ball Bumped Bare Die Sales [BUMPED_CHIP]	Reel, 500	CD-56-1

¹ Z = RoHS Compliant Part.

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